

Kinetic Inductance Detectors

- **INTRINSICALLY MULTIPLEXABLE**
- **FABRICATION MUCH EASIER**
- **NEW CONCEPT, STILL TO BE STUDIED**

Which applications ?

- | | |
|-------------------------------------|-------------------|
| - mm and sub-mm wave astronomy | DEFINITELY YES |
| - high energy resolution X-rays | PROBABLY NOT |
| - Optical Astronomy | POSSIBLE, but ... |
| - Dark matter and Neutrino | PROBABLY YES |
| -many others to be proposed... | SURE |

Multiplexing schemes

TIME-DOMAIN

Reading-out one detector at a time.

Examples:

- TES (NIST, APC, Germany)
- Semiconductor bolometers
e.g. CEA for Herschel, Néel for NbSi bolometers

ADVANTAGE: simple

DRAWBACK: few, «slow» pixels*

* Remark valid for LTDs only.

(RADIO) FREQUENCY-DOMAIN

Each detector is associated to a frequency f_i .

f_i ; $i=1,2 \dots N$

Each detector has a Δf available (speed).

→ $N \cdot \Delta f$ is roughly the total band we need.

If N and/or Δf is large (usually desired)

→ Need enough space in frequency domain

→ **RF (pixels broadcast)**

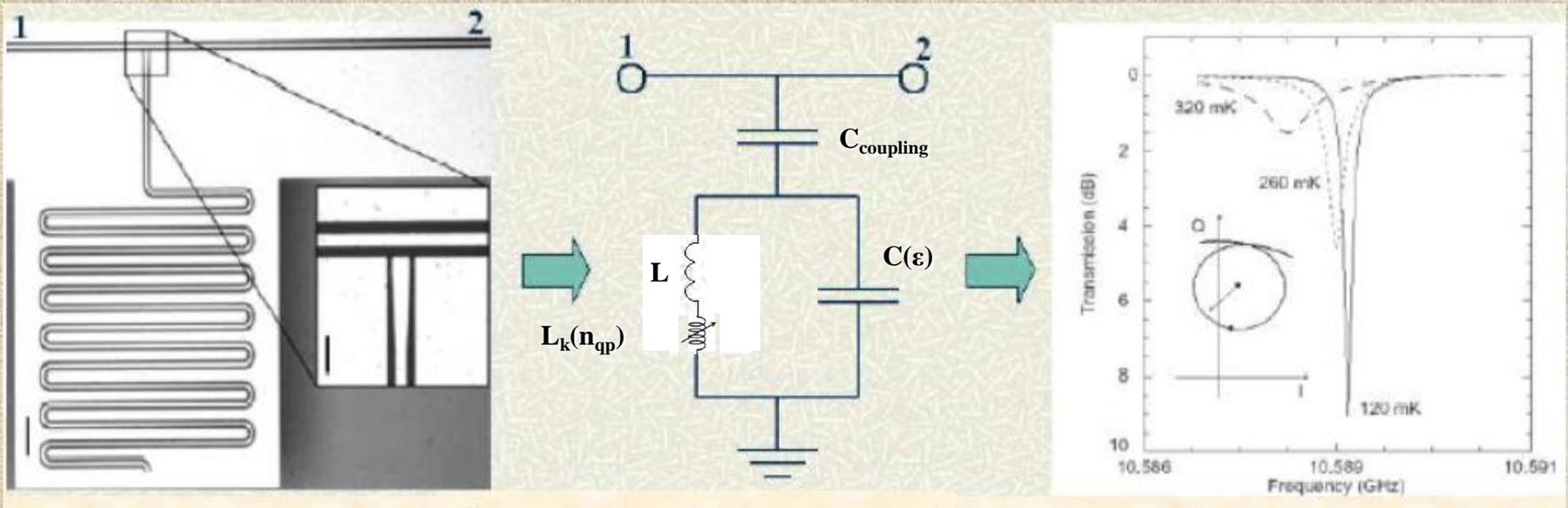
e.g. $\Delta f = 1\text{MHz}$, $N=1000 \rightarrow 1\text{GHz}$

ADVANTAGE: many, «fast» pixels

DRAWBACK: ?? Let see ??

High Quality Factor Superconducting Resonators

Kinetic Inductance “Detectors”



MULTIPLEXING FIRST !! RF and f-DOMAIN IS THE BEST OPTION. HOW ?

High MUX factor \rightarrow closely packed in $f \rightarrow$

Resonating structures (high Q) \rightarrow

SUPERCONDUCTORS $T \ll T_c$

Parallel C-coupled resonators \equiv MUX scheme \neq Detectors

Can the MUX scheme be transformed into a Detector ? DEFINITELY !

Some typical dielectric waveguides

Comparison

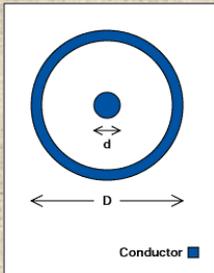
← No cut-off frequency →

← homogenous inhomogenous →
 ⇒ no dispersion ⇒ ϵ_{eff} and Z_0 depend
 ⇒ no impedance mismatch on frequency ⇒ dispersion and impedance mismatch

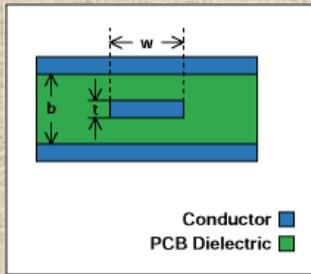
← Good isolation between adjacent traces can be achieved

Can be fabricated using planar technology →

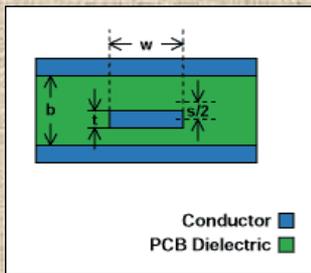
Are sensitive on the material above →



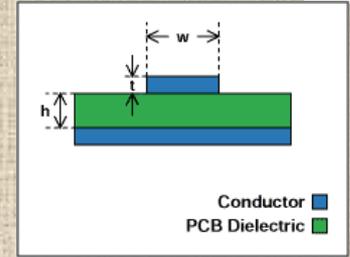
Coaxial line



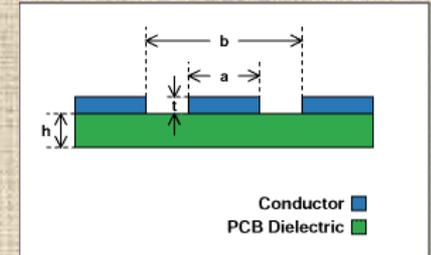
Centred stripline



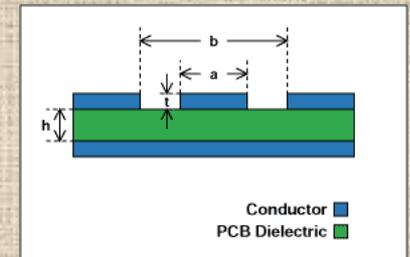
Off-centred stripline



Microstrip line

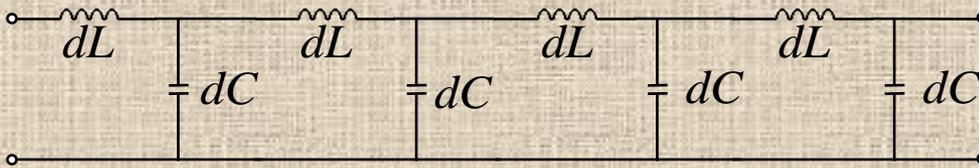


Coplanar waveguide



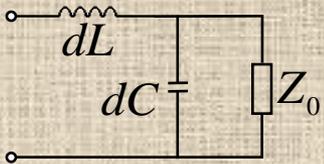
Grounded coplanar waveguide

Each transmission line can be represented as series inductances and shunted capacitances



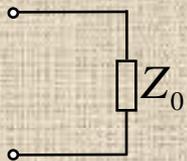
$$dL = \mathcal{L} dl, [\mathcal{L}] = \frac{H}{m}$$

$$dC = \mathcal{C} dl, [\mathcal{C}] = \frac{F}{m}$$



From this approximation we can extract some useful relations:

Characteristic impedance: $Z_0 = i\omega dL + \frac{1}{i\omega dC + 1/Z_0} \Rightarrow Z_0 = \sqrt{\frac{\mathcal{L}}{\mathcal{C}}}$



Phase velocity:

$$v_{phase} = \frac{1}{\sqrt{\mathcal{L}\mathcal{C}}} = \frac{c}{\sqrt{\epsilon_{eff}}} = \frac{\omega}{k}$$

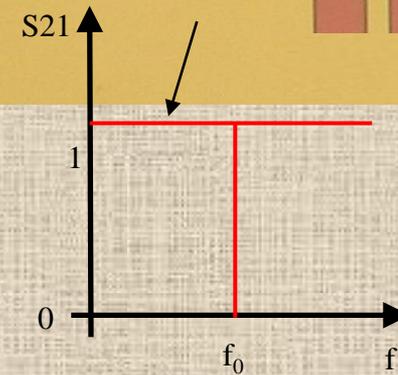
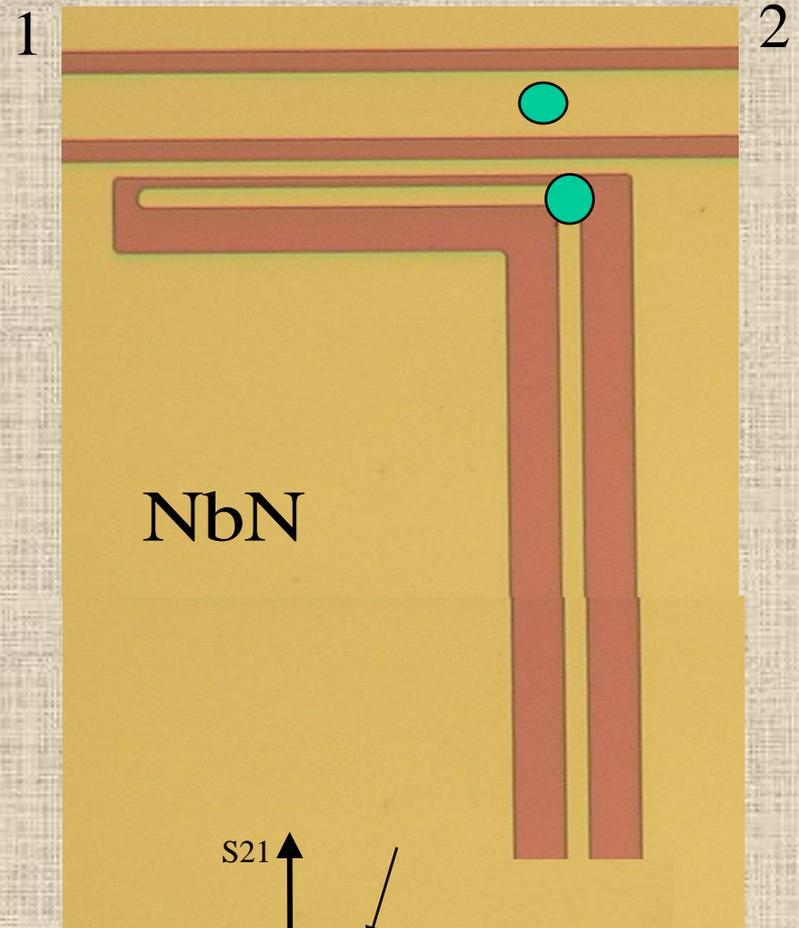
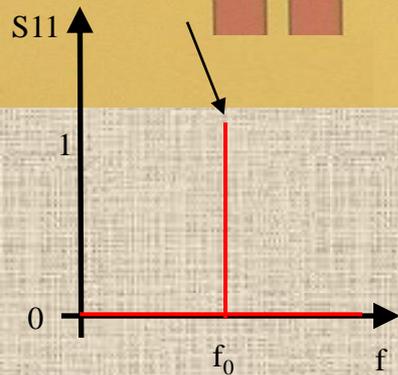
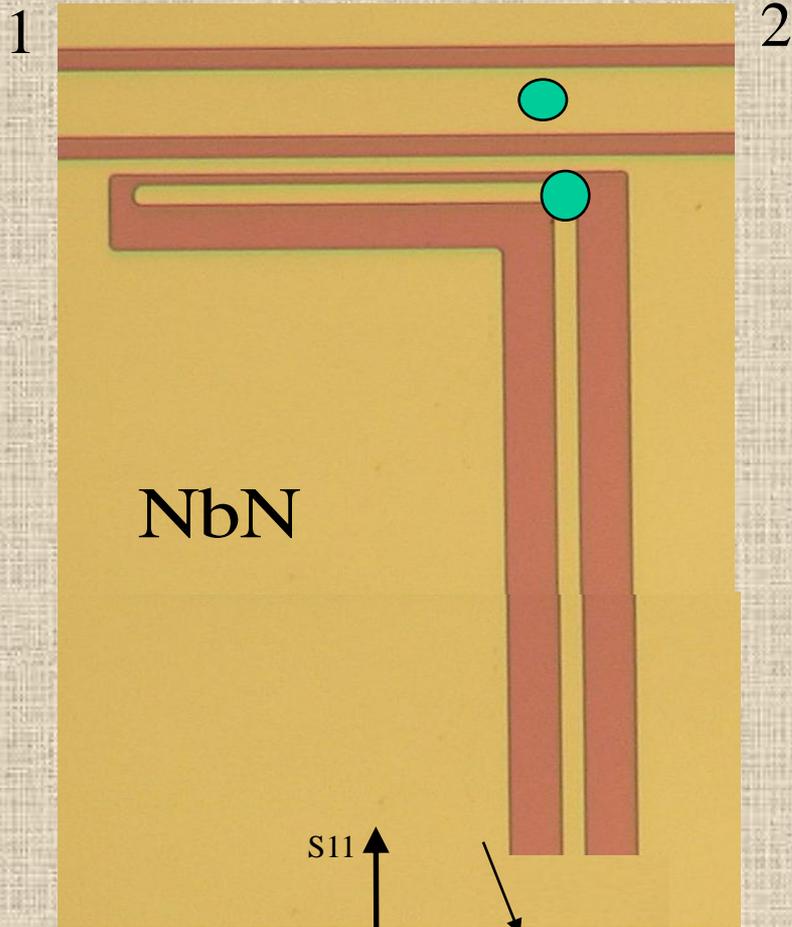
Some typical values:

$$\mathcal{L} \approx 250 \text{ nH/m} \text{ and } \mathcal{C} \approx 100 \text{ pF/m}$$

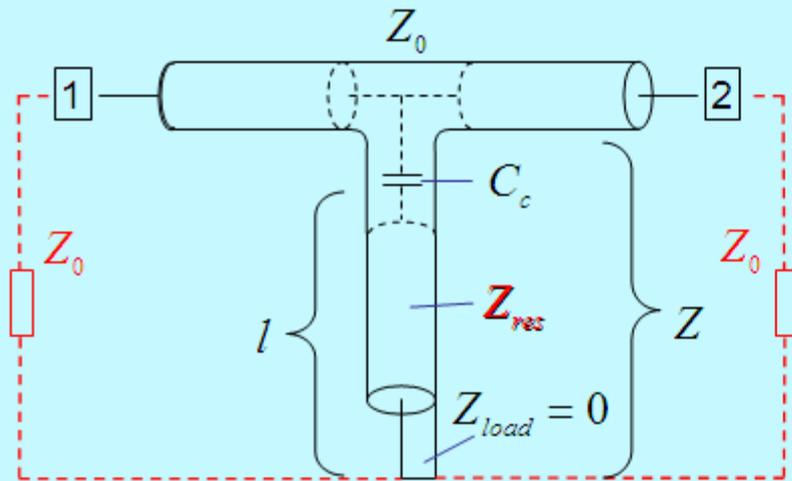
$$Z_0 = 50 \Omega; v_{phase} = 2 \cdot 10^8 = 0,7c$$

ON RESONANCE: $|S_{11}|=1$

OFF RESONANCE: $|S_{21}|=1$



High-Q resonators



Dissipation of energy due to:

- Loss in the resonating transmission line (dielectric and superconductor) → Q_i
- The resistance Z_0 at both ports → Q_c

$$S_{21} = \frac{S_0 + 2iQ_0\delta x}{1 + 2iQ_0\delta x}$$

$$Q_0 = \frac{Q_c Q_i}{Q_c + Q_i}$$

$$S_0 = \frac{Q_c}{Q_c + Q_i}$$

Total (measured) Q factor and minimum transmission.

$$f_0 \cong \frac{1}{4l \sqrt{(L_K + L_G)C(\epsilon)}},$$

$\bullet \quad \bullet$
 L_{tot}

Resonating frequency

Linear dependence on l . Assuming:

$$\Delta l/l = 10^{-4} \quad (\Delta l = 2\mu\text{m}; l = 20\text{mm}) = \Delta f/f$$

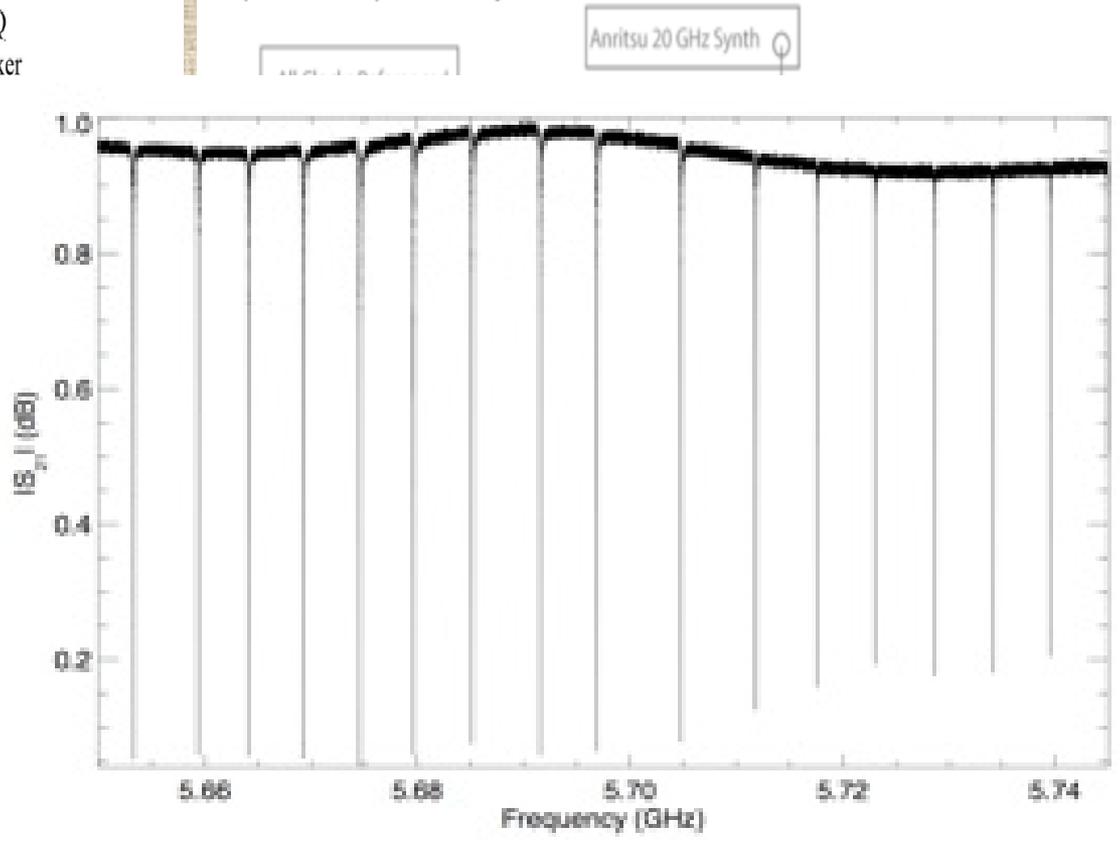
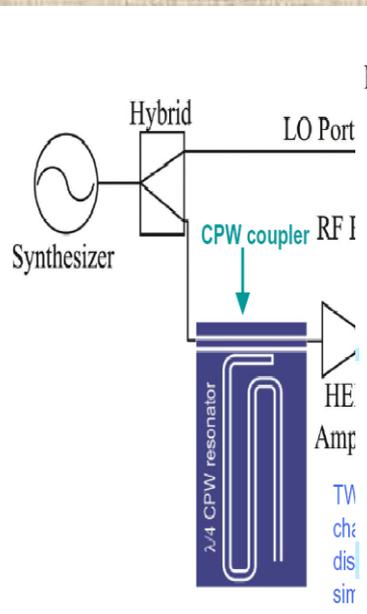
$$\rightarrow \Delta f = 0.4 \text{ MHz @ } f = 4 \text{ GHz}$$

Main design tool: SONNET

The screenshot displays the Sonnet software interface, which is used for electromagnetic simulation. It consists of several windows:

- Top Window (xgeom 9.52):** Shows a schematic of a resonant coupler structure. The structure is a vertical rectangular shape with a horizontal bar across the top. Two ports are labeled '1' and '2'. The structure is filled with a red cross-hatch pattern representing a metal mesh.
- Right Window (emgraph 9.52):** A Cartesian Plot showing the magnitude response in dB versus Frequency in GHz. The plot has two curves: a blue curve for DB[S11] and a magenta curve for DB[S12]. The blue curve shows a resonance dip at approximately 8.678 GHz, while the magenta curve shows a resonance peak at approximately 8.679 GHz. The y-axis ranges from -35 dB to 0 dB, and the x-axis ranges from 8.678 GHz to 8.68 GHz.
- Bottom Window (Metal Types):** A dialog box for defining metal types. It lists three metal types:
 - Lossless: Cnd= INF
 - Nb: Rdc= 5e-8 Rrf= 0 Xdc= 0 Ls= 0.06
 - NbN: Rdc= 0 Rrf= 0 Xdc= 0 Ls= 1.463
 The 'Metal for New Polygons' dropdown is set to 'Lossless'. Buttons for 'Add...', 'Edit...', 'Remove', and 'Library...' are visible.

Readout



PC with 2 TB
Raid 5 Array
for data
processing and
storage

Xilinx SX95T
RF Engines ChannelCore
144 channel complex IQ
demodulation

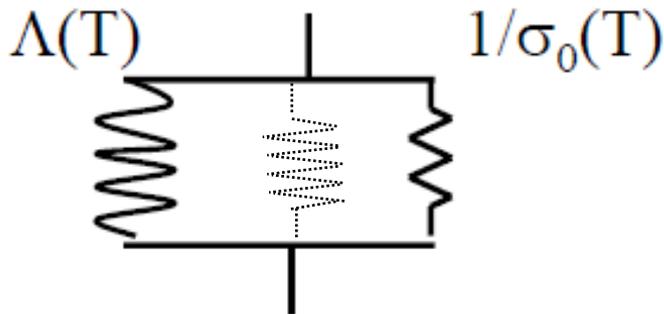
Analog readout
(1 mixer; no ...)

(UX)

KINETIC INDUCTANCE DETECTORS

Two-fluid model

Supercarriers Additional losses (Q_i) Normal carriers



Supercarriers

Normal carriers

Actually $N_n \sim e^{-\Delta/kT}$



Cooper pairs



« quasi-particles »
(un-paired electrons)

$$Z_{\text{tot}} = 1/(1/R_n + 1/i\omega L + 1/R_{\text{losses}})$$

$$Z_{\text{tot}} \rightarrow 0 \text{ if (and only if due to } R_{\text{losses}}) \omega \rightarrow 0$$

Model: two separate populations of current carriers co-exist in a finite T superconductor. So each superconducting strip is seen as a parallel of « normal resistance » (quasi-particles) and « super-inductance » (Cooper pairs).

In AC the « super-inductance » is no longer a zero impedance shunt

→ Quasi-particles losses for any finite T superconductor in AC.

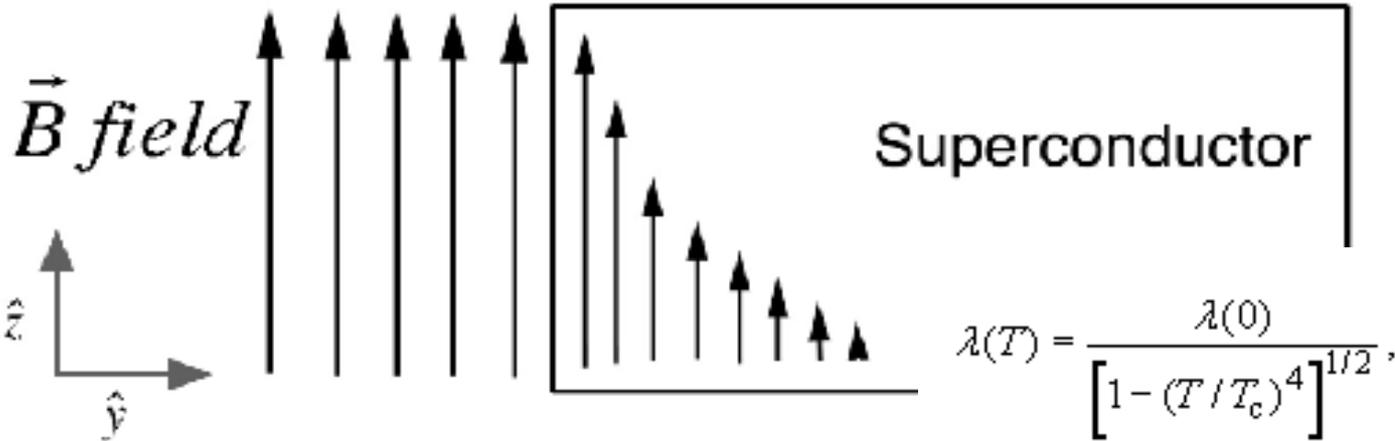
→ Additional losses (e.g. dielectrics) always present to limit Q_i even at $T \rightarrow 0^2$

Kinetic Inductance: we have a detector !!

When
field
actual
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L_k is



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$$L_K = \frac{\mu_0 \lambda(T)^2}{l} \int \frac{j_s^2}{I^2} dV, \approx \mu_0 \frac{\lambda(T)^2}{l \cdot t}$$

lines), L_k^\dagger depends on the penetration length λ , the quasi-particles concentration and so,

finally, the flux of Cooper-pair breaking photons hitting the slab.

→ MUX IS SEEING LIGHT

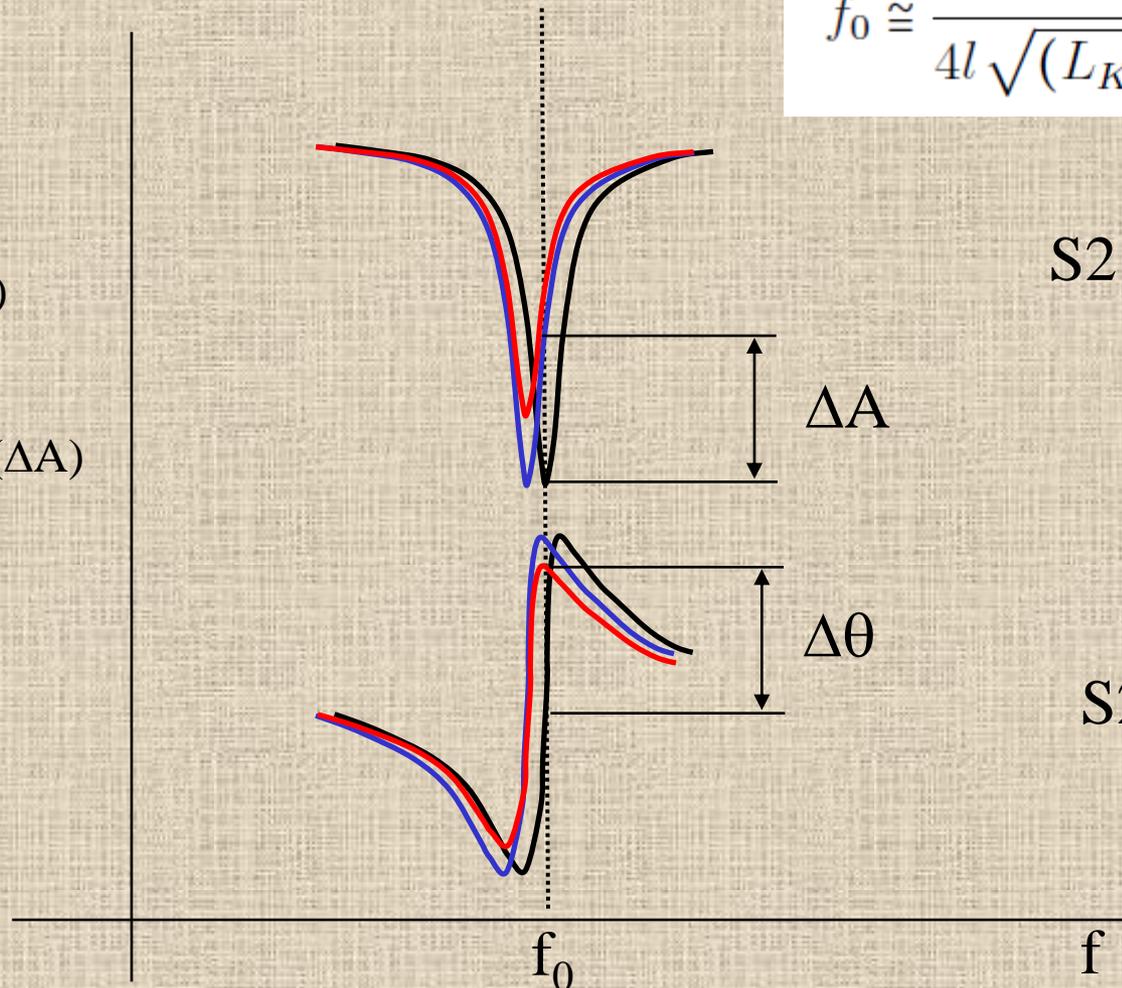
Switch on the light

$$f_0 \cong \frac{1}{4l \sqrt{(L_K + L_G)C(\epsilon)'}}$$

Dark, $T \ll T_c$

Light: increase in L_k
Change in phase ($\Delta\theta$)

Light: increase in R
Change in amplitude (ΔA)



S21 Amplitude

S21 Phase

A superconducting resonator is very sensitive to

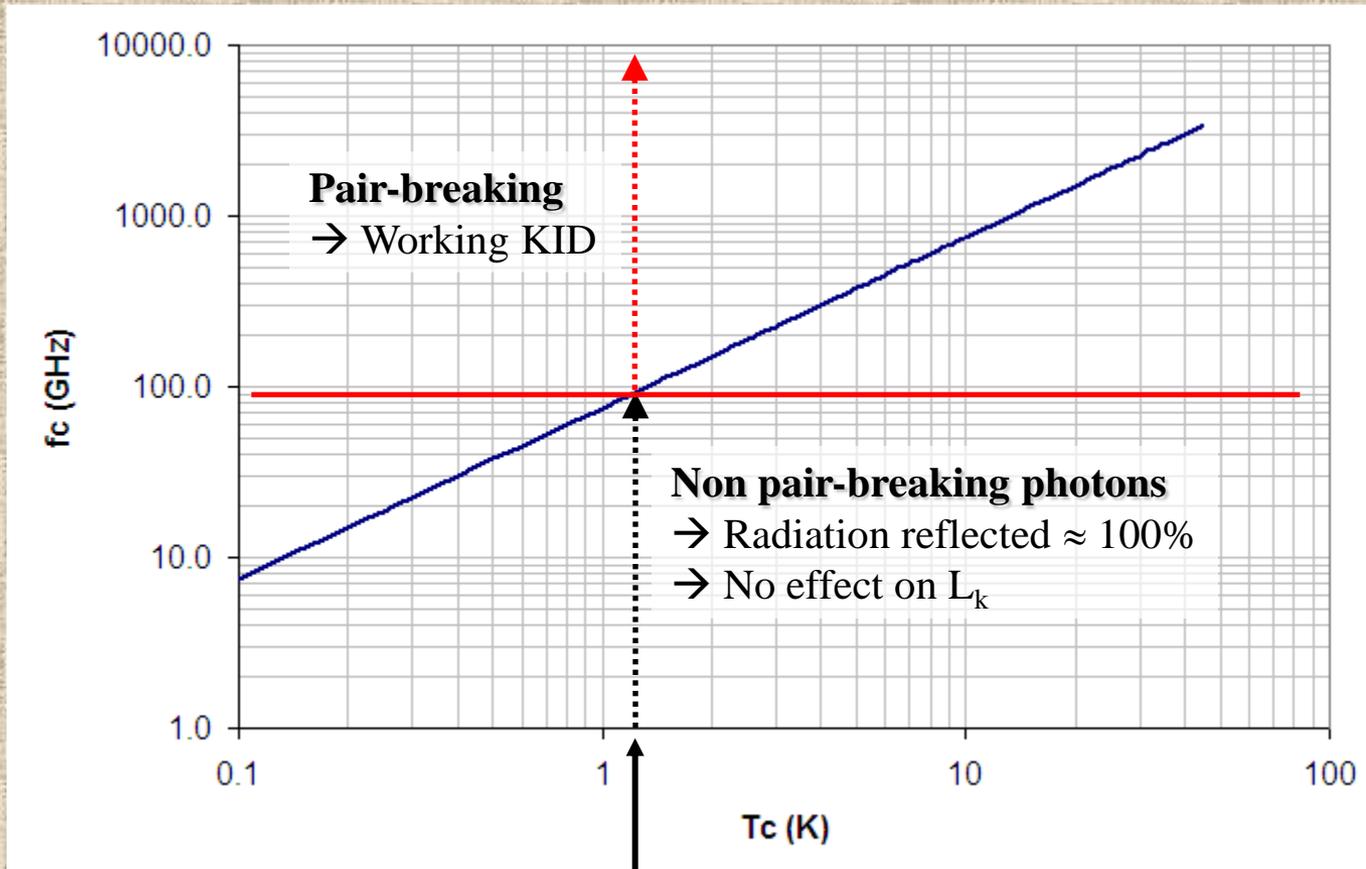
- **the kinetic inductance (strictly speaking KIDs)**

- photon detection (ONLY if incident radiation $h\nu > 2 \cdot E_{\text{gap}} \approx 3.6 \cdot kT_c$)
- temperature dependence (if $T > 0.2 \cdot T_c$... otherwise not so sensitive)
- measurement of the penetration depth λ_L (if $T > 0.2 \cdot T_c$... otherwise frozen at λ_0)

- **the EM environment**

- characterization of materials and the change of ϵ_r with temperature
- measurement of the loss $\tan \delta$ in the surrounding dielectric medium and of the attenuation constant α . Sensitive to losses in the ground plane.
- small changes in the geometry (e.g. nano-mechanics, hydrodynamics)

MATERIALS FOR LOW-ENERGY PHOTON DETECTIONS



e.g. Al

Examples:

Ti $\rightarrow f_c \approx 40\text{GHz}$

Al $\rightarrow f_c \approx 90\text{GHz}$

Re $\rightarrow f_c \approx 130\text{GHz}$

Ta $\rightarrow f_c \approx 340\text{GHz}$

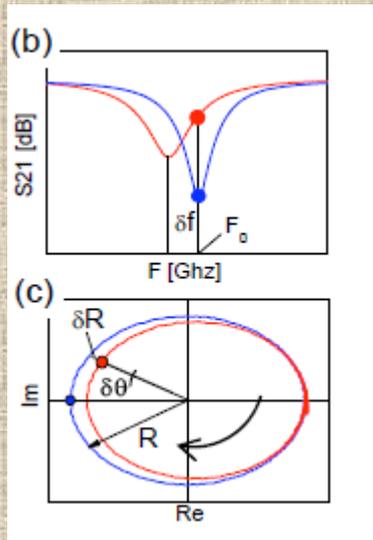
Nb $\rightarrow f_c \approx 700\text{GHz}$

NbN $\rightarrow f_c \approx 1.2\text{THz}$

...

Sensitivity

Two possible measurements on the complex S21 (transmission) plane:
radius (**amplitude**) and azimuth (**phase**).



$$\delta\theta = [2Q_i^2 / (Q_i + Q_e)] (\delta f / f_0) \cong 2 \cdot Q \cdot (\delta f / f)$$

The Q factor and the resonator volume determine directly the sensitivity.

$$\frac{d\theta}{dN_{qp}} = 1.63 \times 10^{-7} \frac{\alpha Q}{V},$$

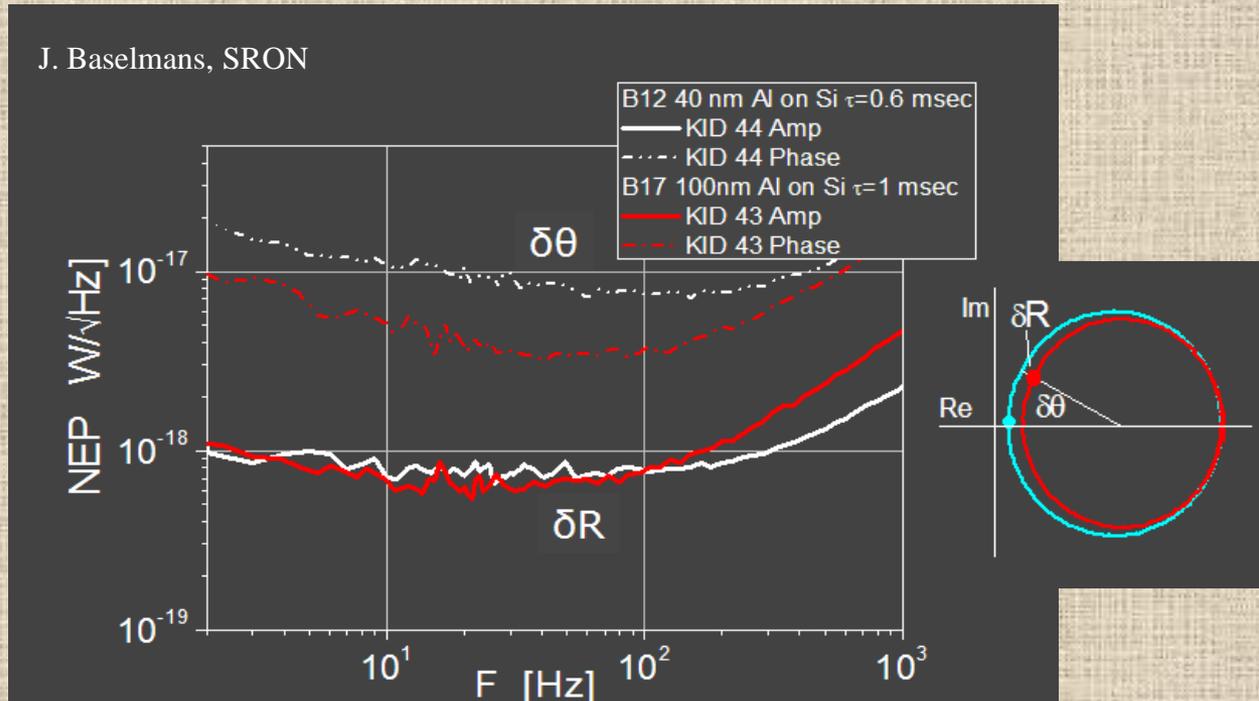
Al

B. Mazin, PhD thesis, Caltech 2004

| Noise | Signal |
|---|--------|
| $\text{NEP} = S_x(\omega) \cdot \left(\frac{\eta \tau_{qp}}{\Delta} \frac{\partial X}{\partial N_{qp}} \right)^{-1} \cdot \sqrt{1 + \omega^2 \tau_{qp}^2}$ | |

→ **high-Q (film quality, design), low V (design, technology), long τ_{qp} (film quality, impurities ... still unclear)**

Noise: phase vs. amplitude read-out

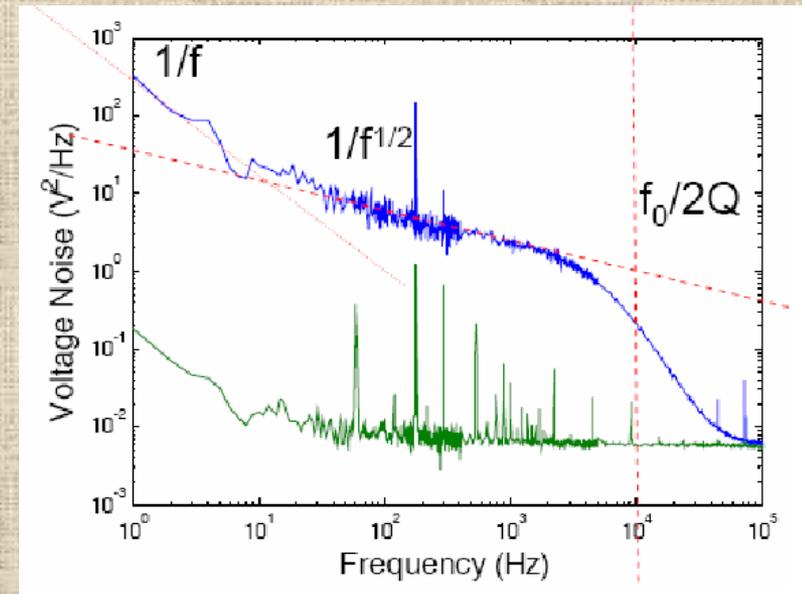
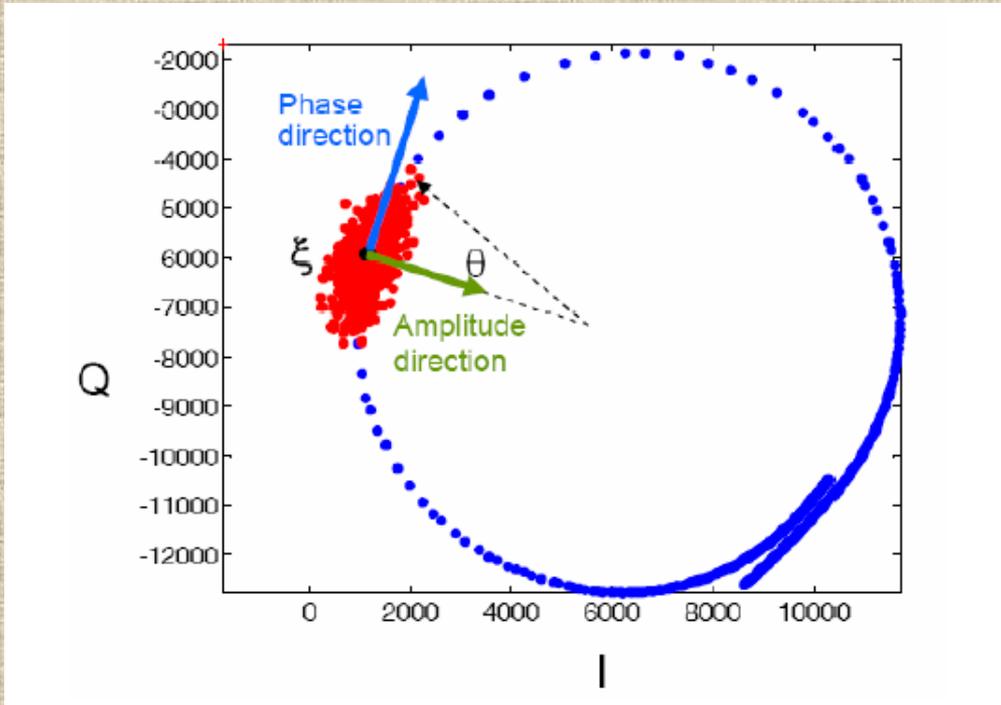


Still some **excess noise** in phase readout.

New results seem to indicate it's originating at the **metal/substrate interface** (but where?).

Amplitude read-out already OK for many applications (e.g. ground-based)
 However, for pushing S/N to the limit it is better to **understand and kill** the phase noise.

Excess Phase Noise



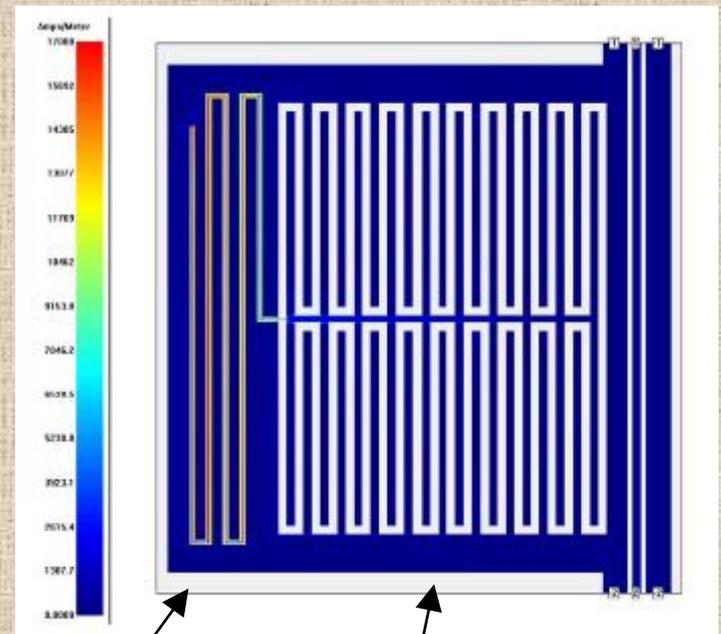
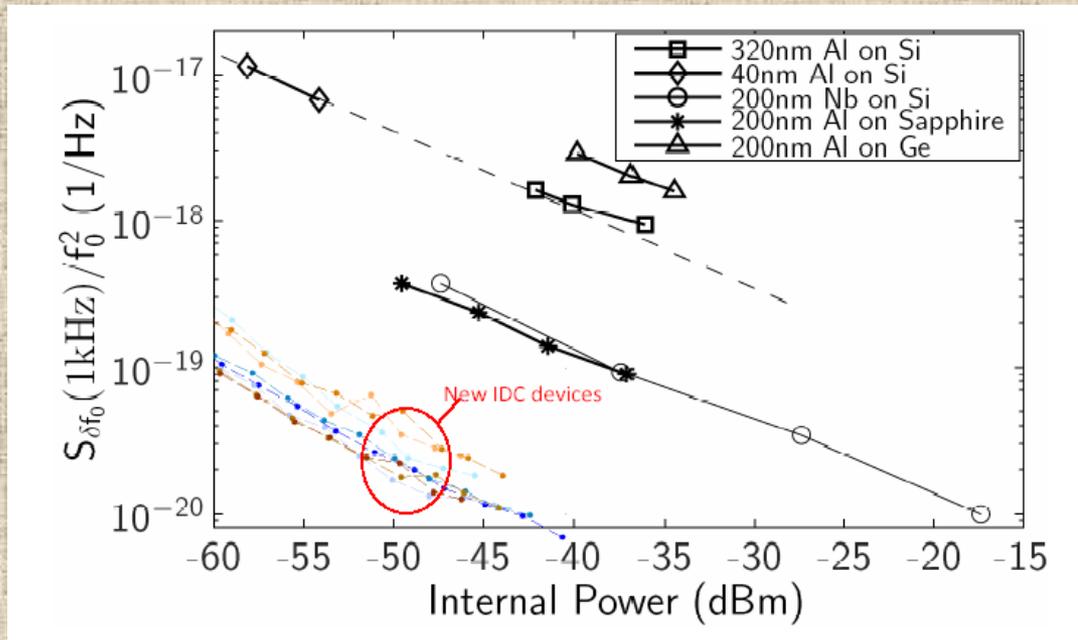
Credit: Omid Noroozian, Caltech (2008)

**A semi-empirical model now available !!
Work mainly carried out in the US.**

Excess Phase Noise: first hints

To reduce **noise**: make **fat** CPW near the coupler end.

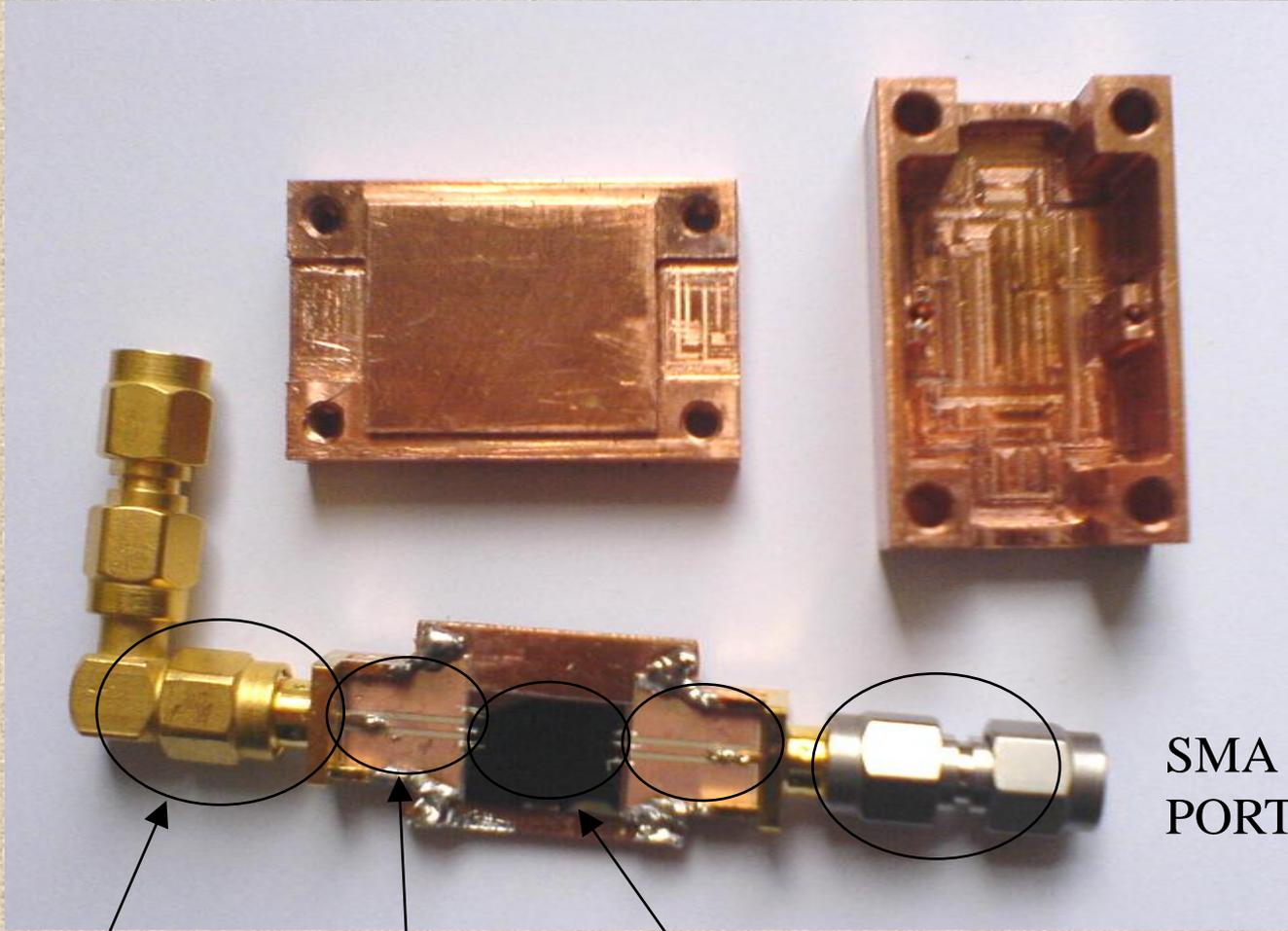
To maintain high **sensitivity**: make **narrow** CPW near short circuit to have higher sensitivity.



Credit: Omid Noroozian, Caltech (2008)

$\lambda/4$ resonator

Interdigitate C



SMA connector
PORT 1

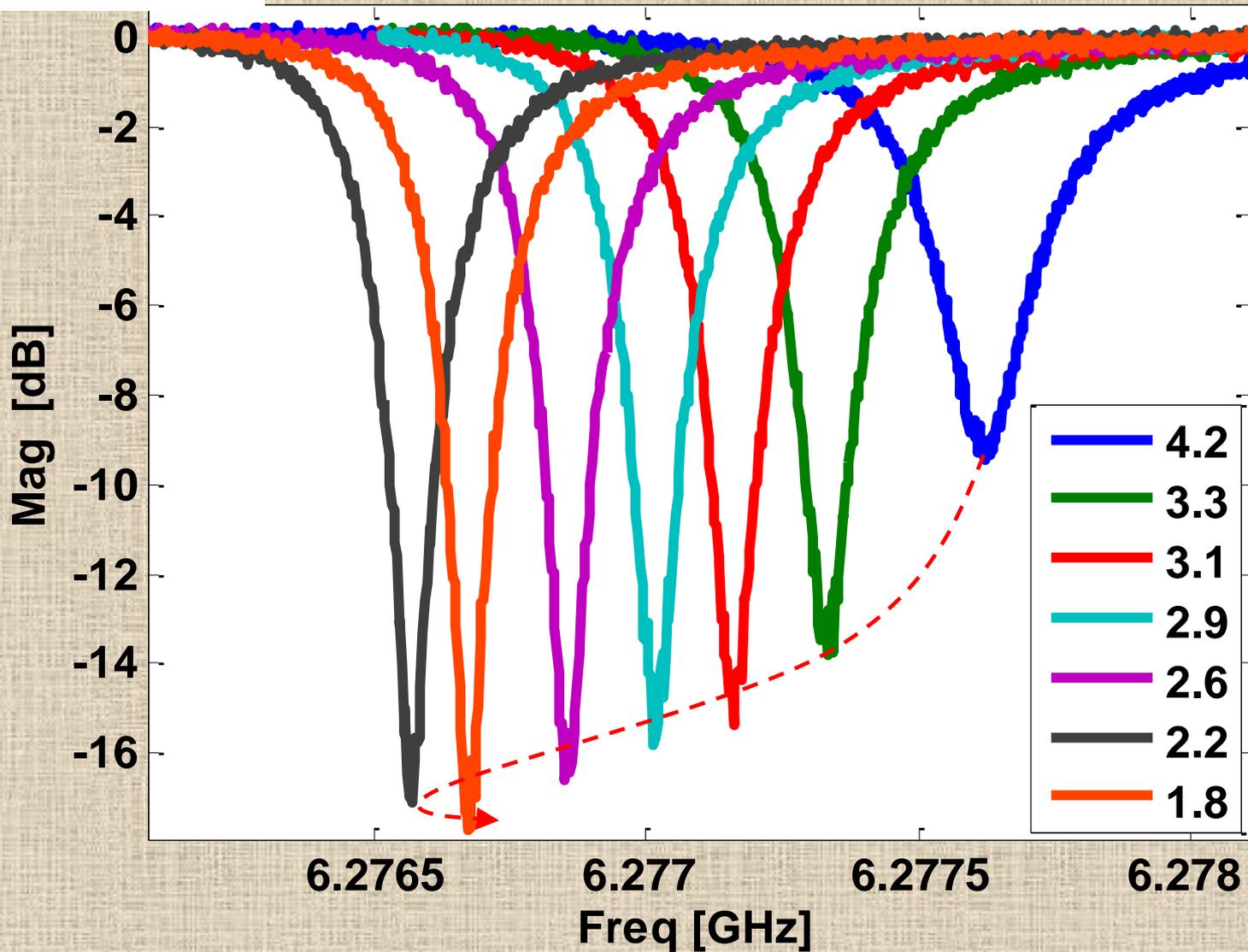
50Ω launcher
(coax-to-planar transition)

Chip with many resonators

SMA connector
PORT 2

$$f_0 \cong \frac{1}{4l\sqrt{(L_K + L_G)C(\epsilon)}}$$

Mag vs. Freq, in dependence of Temp

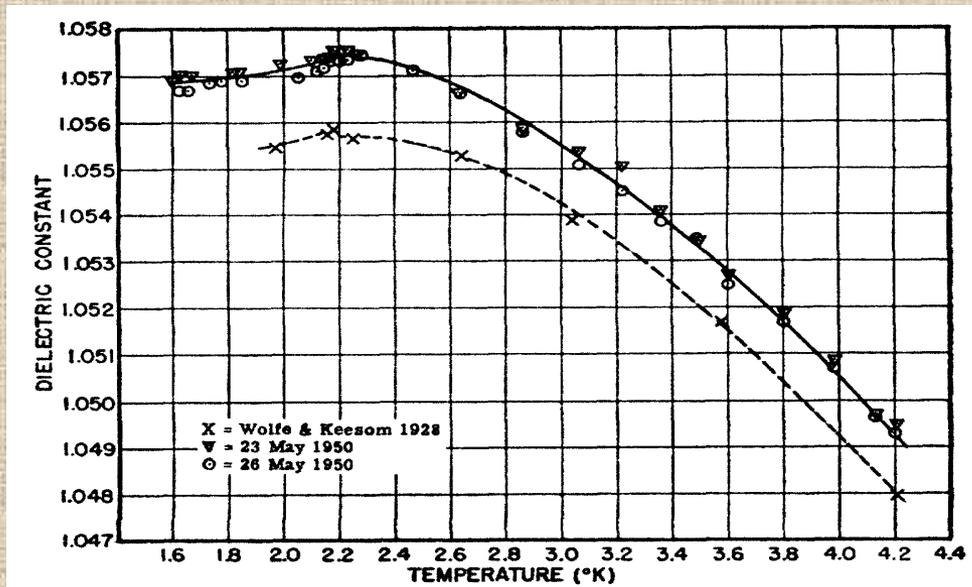


The amplitude and the different quality factors don't show anything special.

But the frequency points clearly an effect.

Solution: Let's see what people did in 1950 (Phys. Rev., Vol.80 Nr.8 Page 89,1950)

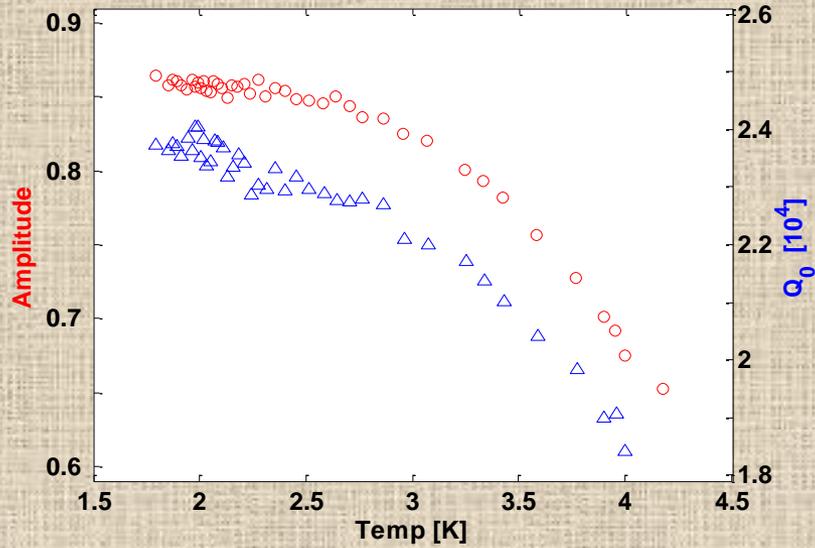
The Dielectric Constant of Liquid Helium



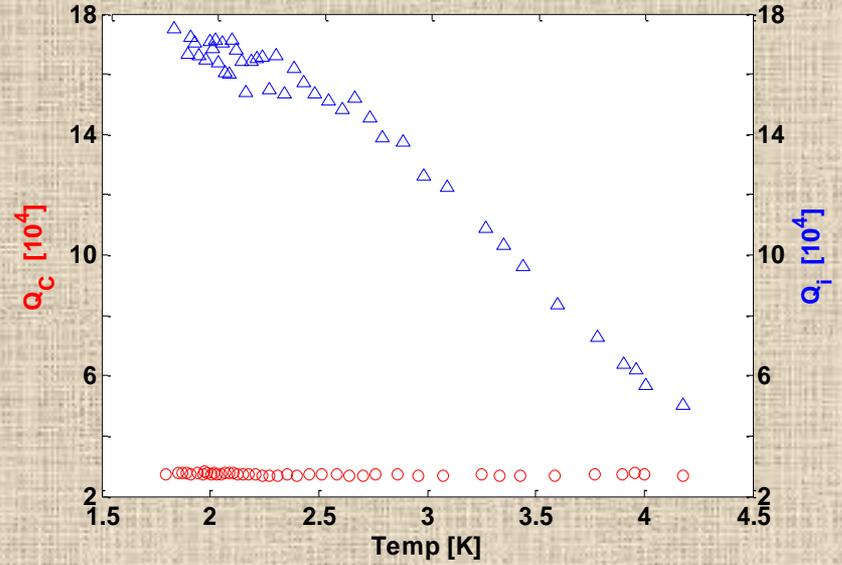
- During decreasing temperature ϵ increases
- At 2.17 K helium gets superfluid \rightarrow first order phase transition
- Below this critical temperature ϵ starts to decrease

Resonances on Sapphire (1.8 K < T < 4.2 K)

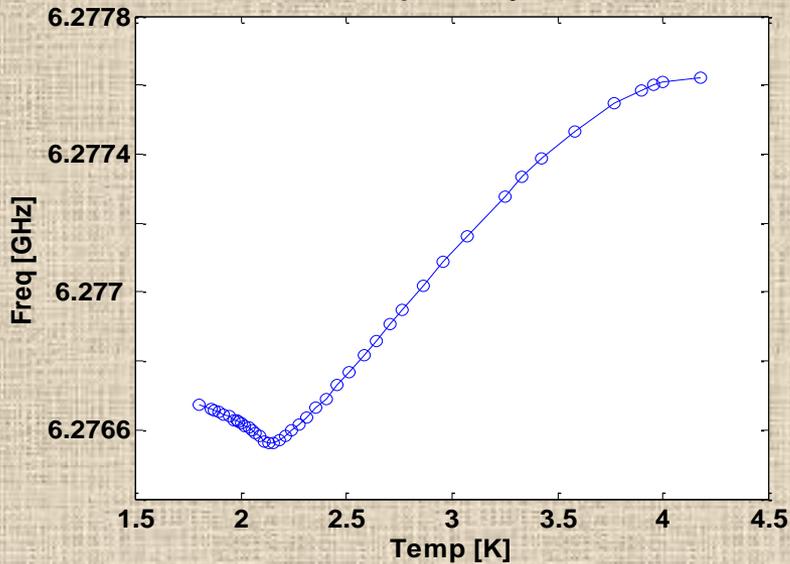
Amplitude, Q - factor



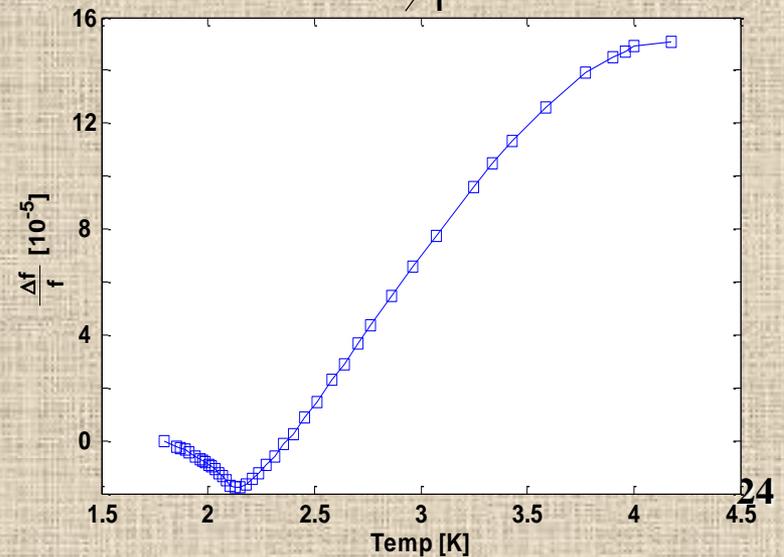
Q_C, Q_i



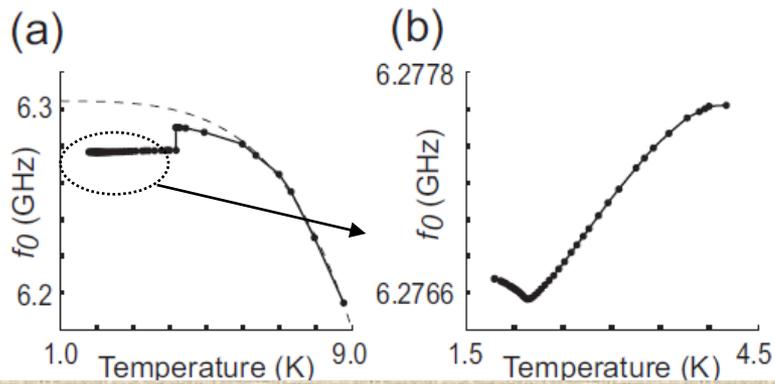
Frequency



$\frac{\Delta f}{f}$



Helium Detectors (Grenoble)

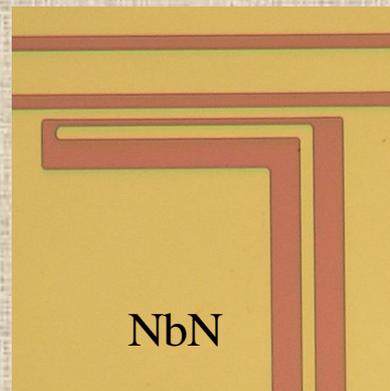


- Designed, fabricated and tested high-Q (10^5) NbN resonators
- Demonstrated a highly sensitive LHe detector for hydrodynamics applications (**APPLIED PHYSICS LETTERS 93, 134102**)
- In progress ..

$$\delta\epsilon_{He}^{min} \cong 2\epsilon \sqrt{\frac{k_B T_n}{2P_s}} \left(\frac{Q_i + Q_e}{Q_i^2} \right)$$

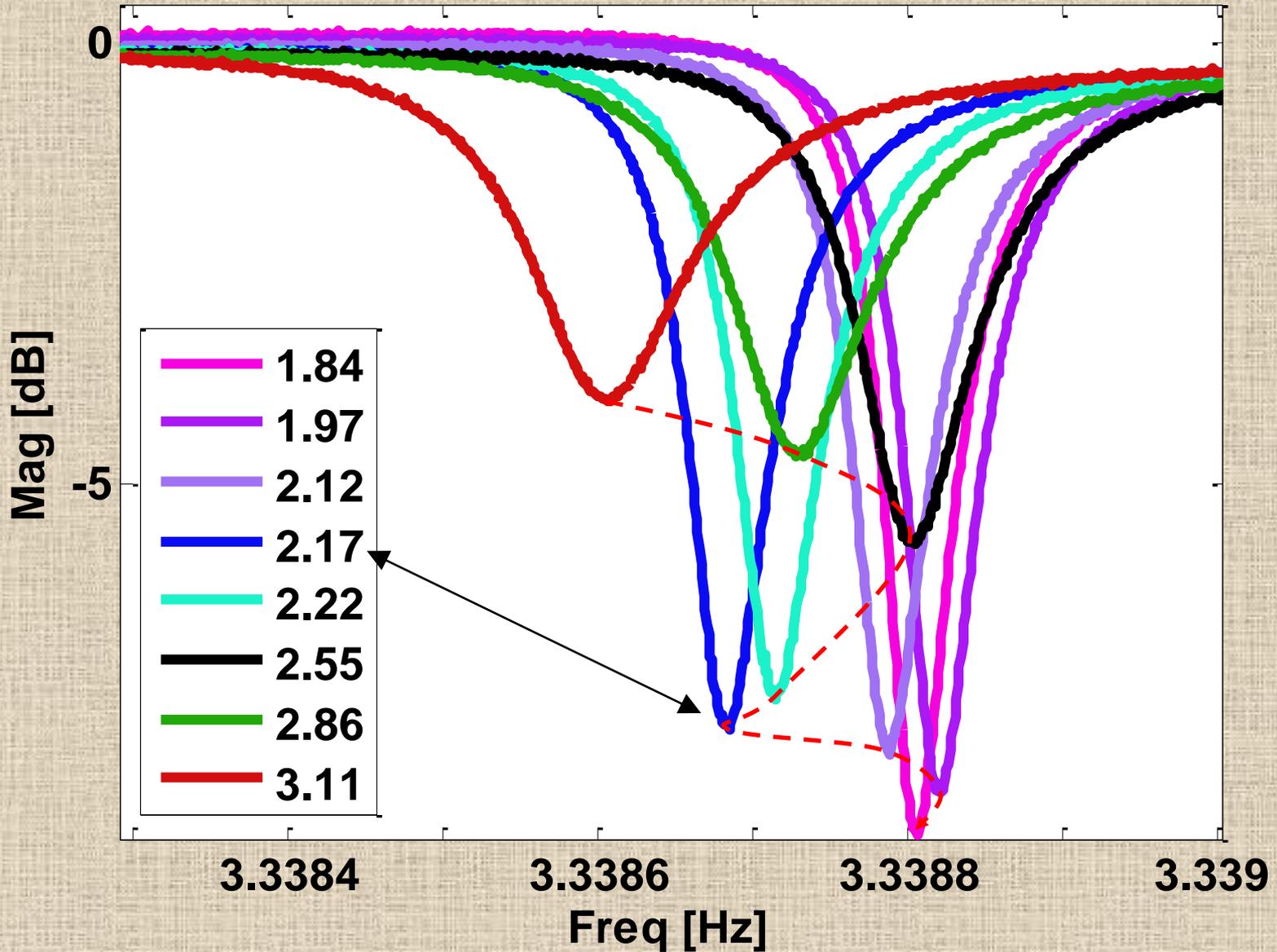
Achievable sensitivity:

$$10^{-10} \epsilon_0 / \text{Hz}^{1/2} \quad !!$$



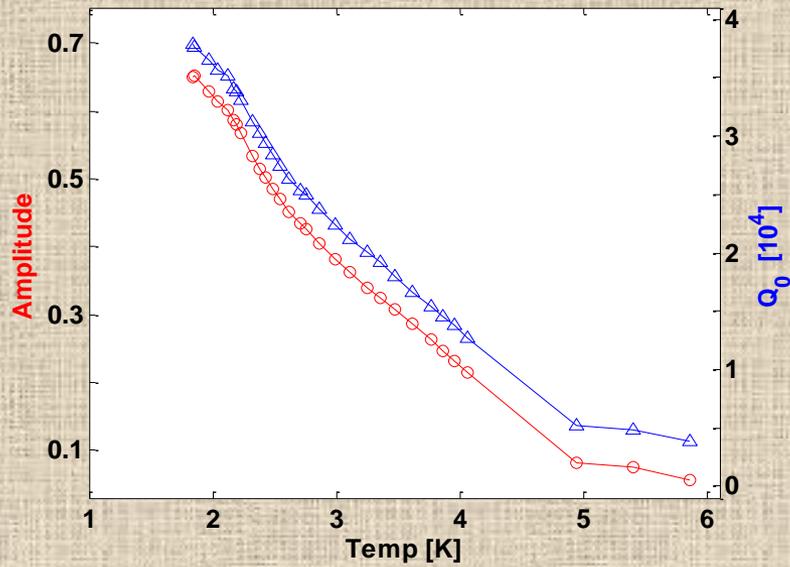
NbN resonators on Si+SiO₂ (1.8 K < T < 4.2 K)

Mag vs. Freq, in dependence of Temp

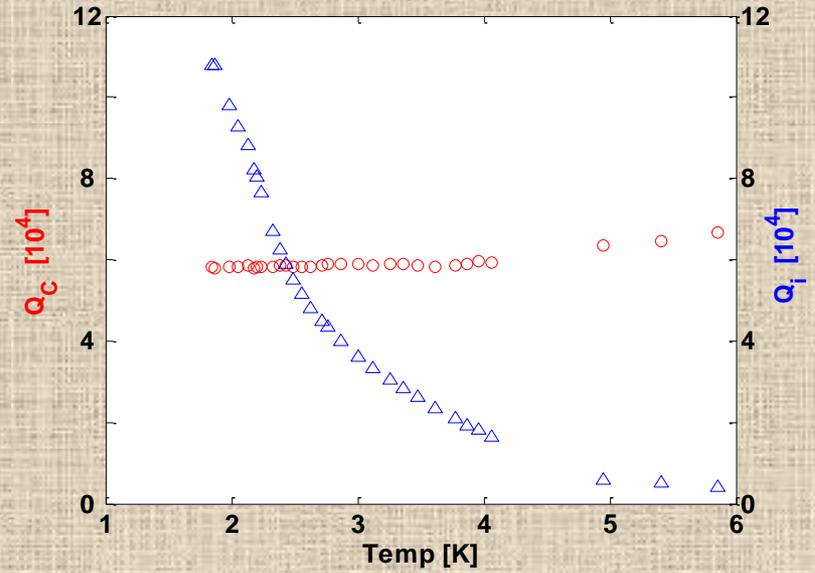


Resonances on Si/SiO₂ in dependence of temperature

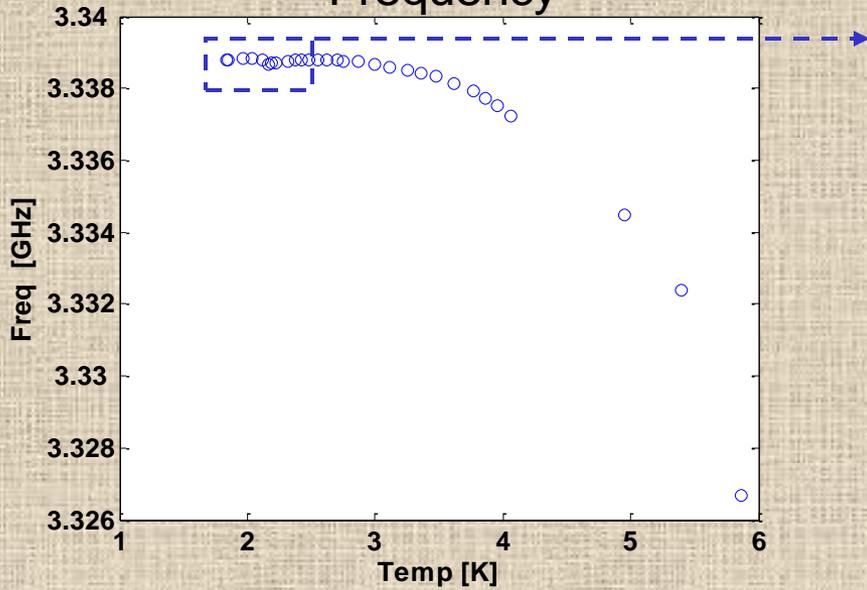
Amplitude, Q - factor



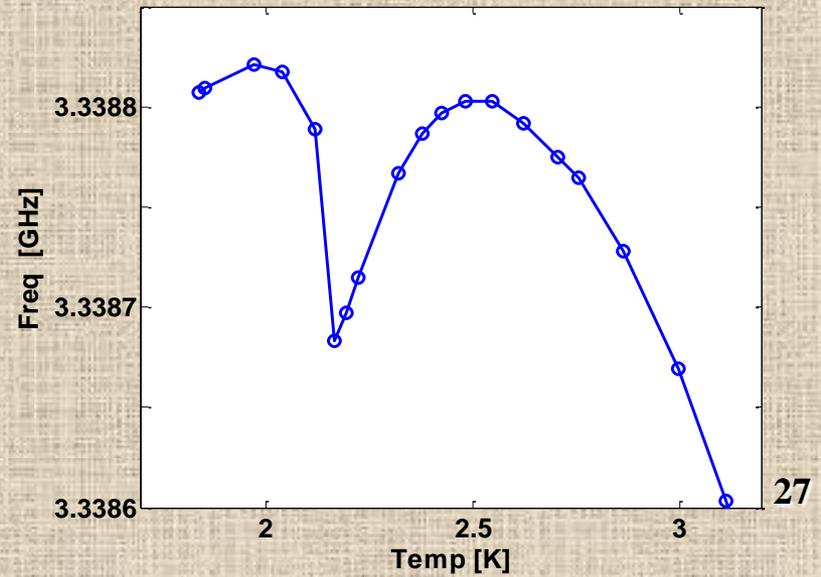
Q_C, Q_i



Frequency



Frequency



Materials for MKIDs

| | | |
|-------|-----------------------|---|
| - Al | $T_c = 1.2 \text{ K}$ | mm-wave detection (down to 90GHz) |
| - Ti | $T_c = 0.5 \text{ K}$ | mm-wave (down to 40GHz) |
| - Nb | $T_c = 9 \text{ K}$ | THz detection (from 700GHz), transmission lines, filters |
| - NbN | $T_c = 16 \text{ K}$ | Hydrodynamics, THz |
| - Re | $T_c = 1.7 \text{ K}$ | Neutrino mass, mm-wave detection, X-rays |
| - Ta | $T_c = 4.5 \text{ K}$ | X-rays absorber, visible-NIR single photons |

- SPUTTERING or UHV EVAPORATION
- BETTER IF HEPITAXIALS (substrate matching)

Bolometers vs KIDs

Bolometers:

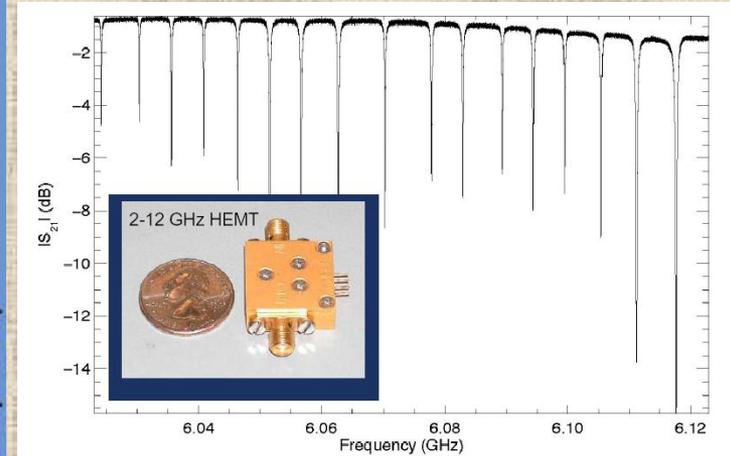
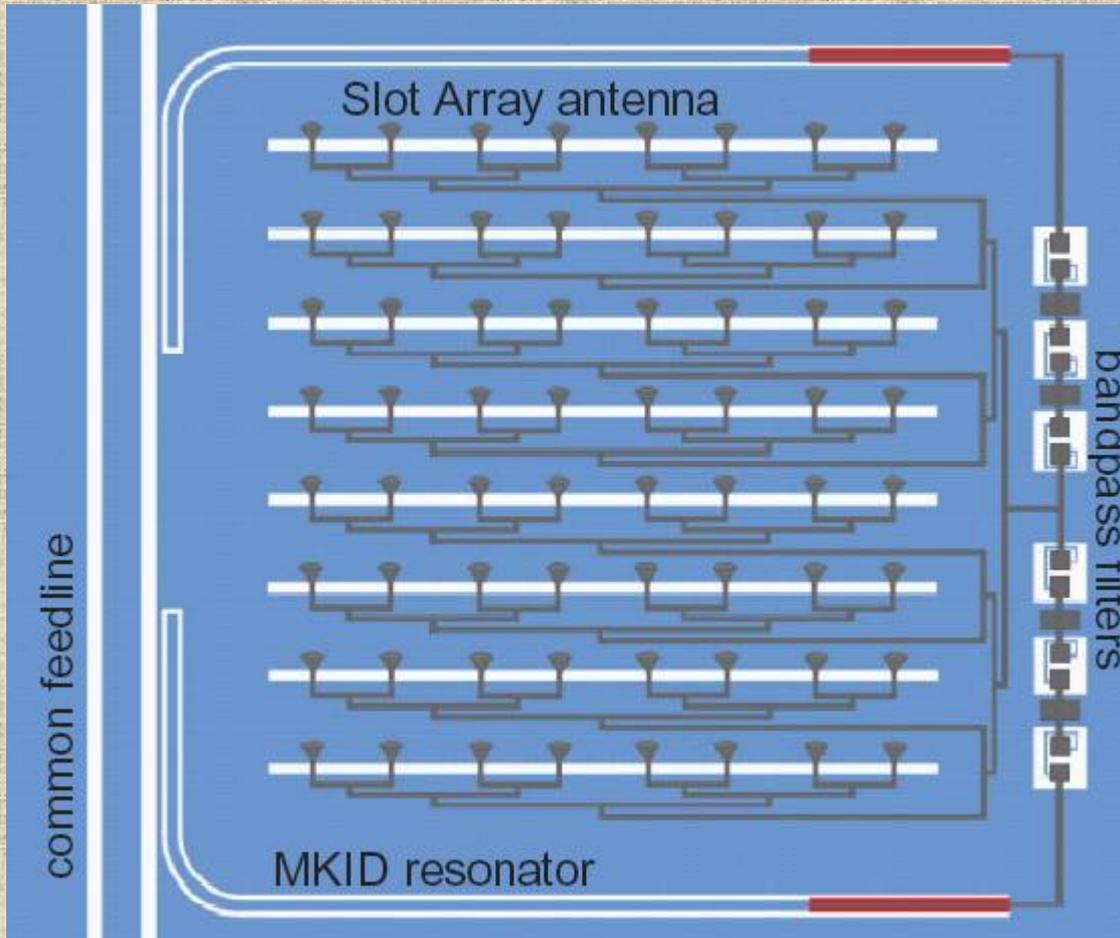
- Sensitivity depends on thermal decoupling, so it's **IN PRINCIPLE** adjustable
- **Easier to decouple the absorber from the thermistor (particularly important for higher energy spectroscopy applications)**
- **Big arrays made by weak multiplexing and brute force**
- **Complicated fabrication processes**
- **Heavy, expensive, complicated low-T electronics (SQUIDs, JFETs, MUX ...)**
-

KIDs:

- Sensitivity is limited by quasi-particles lifetime (not yet under control)
- Quasi-particles transport still difficult (see STJs)
- **Better suited for giant arrays**
- **Easier to fabricate, extremely robust**
- **Less sensitive to T fluctuations; in principle doesn't requires very low T_{base}**
- **Tumultuous R (big) & D (small) phase ongoing (more fun..)**
-

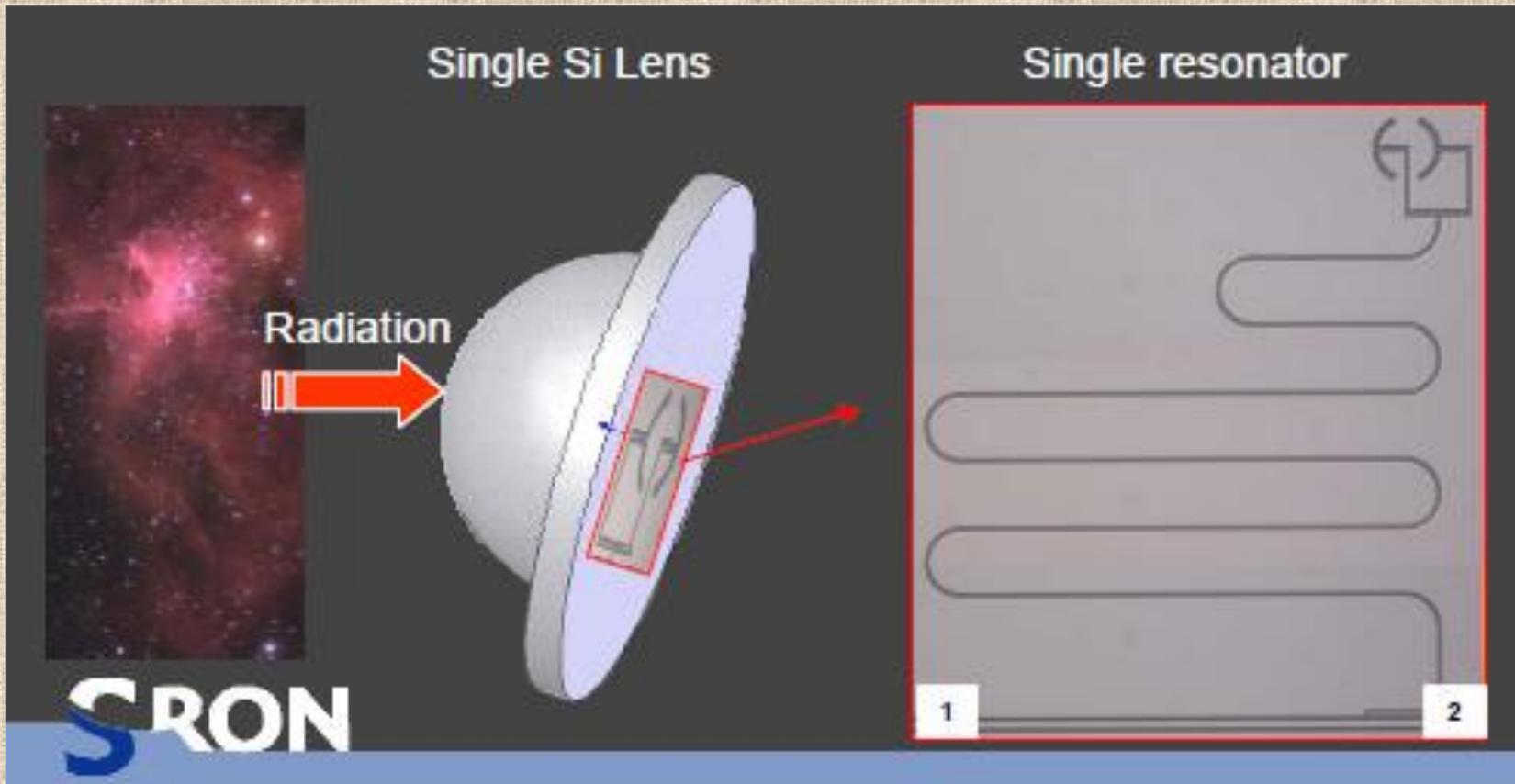
A look around (mm-wave astronomy)

Lossy line over the resonators (USA)



MKIDCAMERA for the Caltech Sub-mm Observatory (CSO)

Antenna IN THE resonator (SRON 2005-)

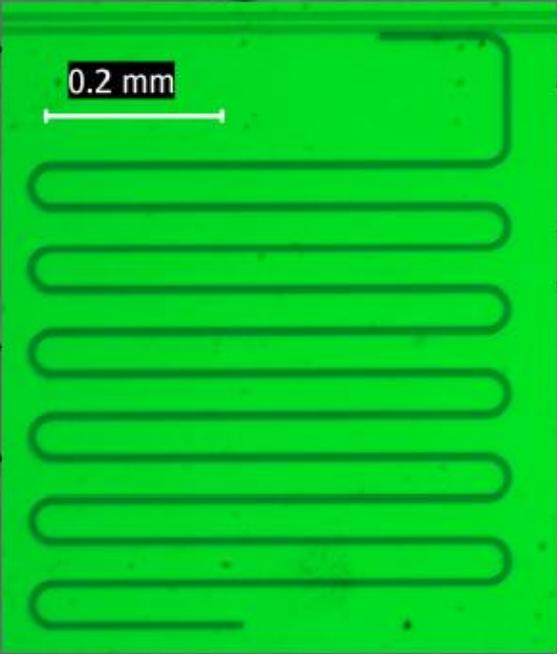


Being developed for SPICA satellite (200 μ m), NIKA-IRAM (2mm) and now also APEX (1mm)

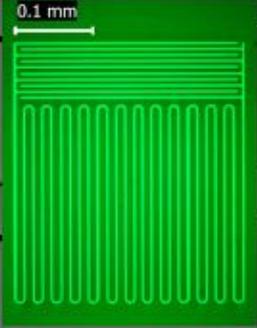
Lumped Element KIDs (Cardiff, Grenoble, Roma)

Distributed vs. Lumped Element Resonators

LEKID



0.2 mm



0.1 mm

LEKID - $f \sim 6$ GHz

Distributed KID - total length 6 mm. $f \sim 6$ GHz

NICA : the Neel IRAM Cardiff Array !!!

A KIDs European Camera for IRAM: the Nèel Iram KIDs Array “NIKA”

Target: Summer 2009

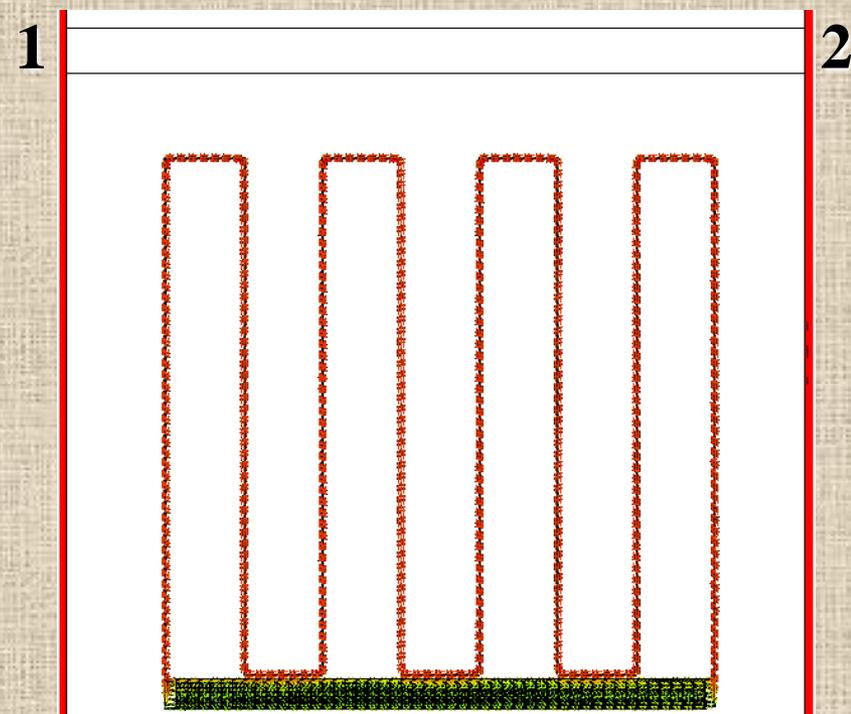
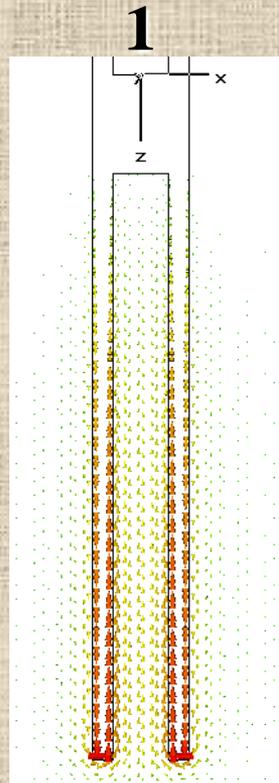
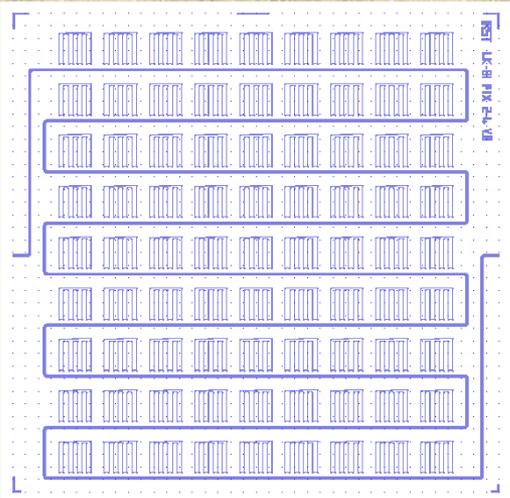
LEKIDs design for IRAM (v.0)

81 pixels LEKIDs array. Now we're at version 1 (CPW and 196 pixels).

Measured preliminary optical NEP around $10^{-15} \text{W/Hz}^{0.5}$, limited by poor film quality and excess phase noise \rightarrow the pixel is absorbing the 2mm radiation !!

$\lambda/4$ (distributed)

Lumped Element



3-D simulation

LEKIDs (1st order) radiation coupling

Incoming radiation



Si substrate $t \approx \lambda_{Si}/4$

Cavity $d \approx \lambda_{vacuum}/4$

Backshort (SC)

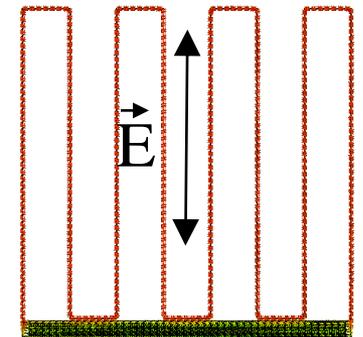
$R_{\square} \approx 1-2\Omega$ for 40nm Al at low-T

The incoming wave is « terminated » with an effective $Z = R_{\square} \cdot D/W_L$ that can be $\approx Z_{Si} = 377/\epsilon_r^{0.5} \approx 100\Omega$

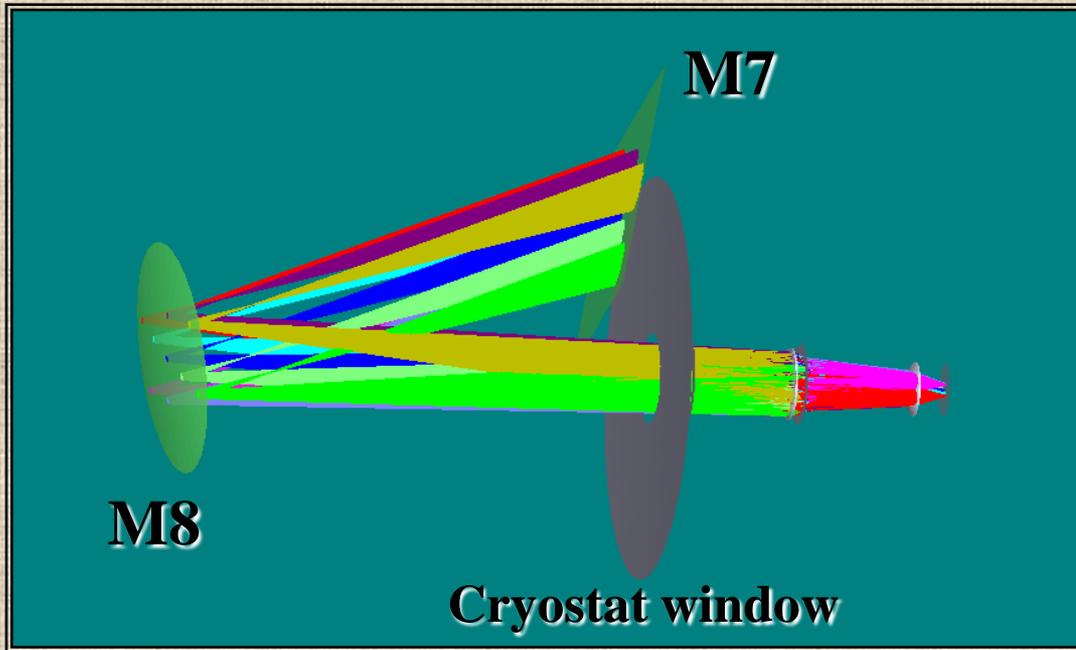
Mean Optical Absorption over 125-170GHz can reach 80% even without an additional AR coating.

D = distance between lines
 W_L = meander line width

$D=280\mu\text{m}$; $W_L=2-4\mu\text{m}$



NIKA goals



On the cryostat (horizontal):

- *M7 (flat)*
 - *M8 (x-y 2nd degree polyn.)*
- at the IRAM focal plane (f/10)

In the cryostat:

- *4 K HDPE lens*
- *100 mK HDPE lens*

Pixel pitch:

1.6 mm ($\lambda = 2.05\text{mm}$, f/1.7 optics \rightarrow Nyquist)

Array dimensions:

32×32 mm²

Number of pixels:

20×20 (2.4×2.4arc-min, pixels spacing 7.2 arc-sec)

Read-outs:

REALLY low-cost FPGA (up to 32 channels)

FFTS (SRON, “best efforts basis”): all the 400 channels

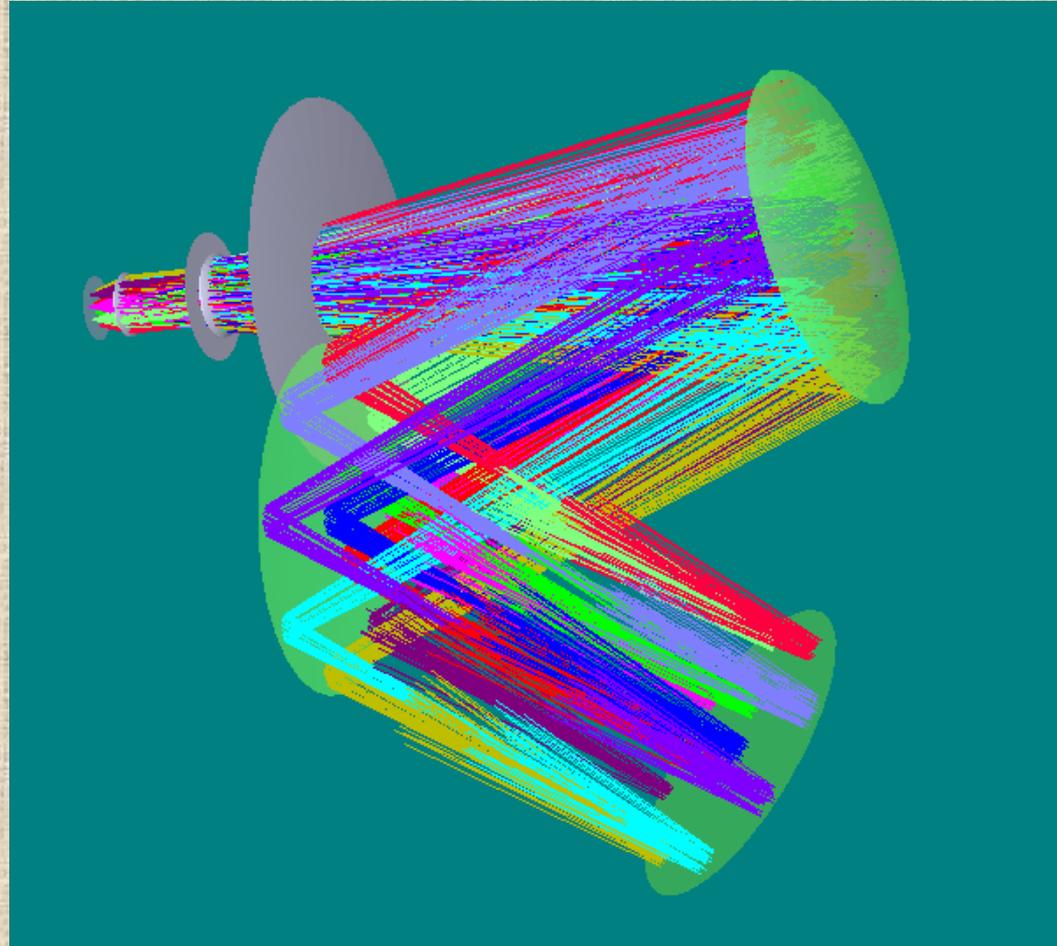
Number of cables from the cryostat: 2 coax (f < 8 GHz), 3 for preamplifier bias.

Perspectives: large FoV

IRAM is going to **increase the 30-m field-of-view**. Several options, with different complexities and costs, are under study. **From 12' downwards**.

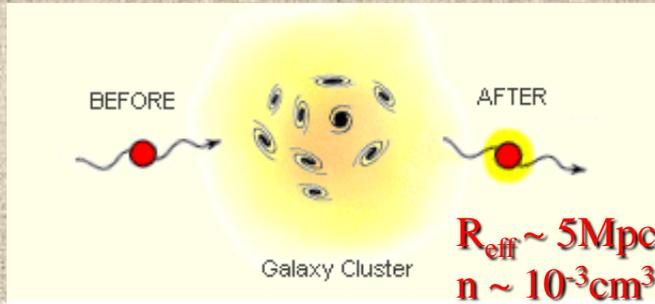
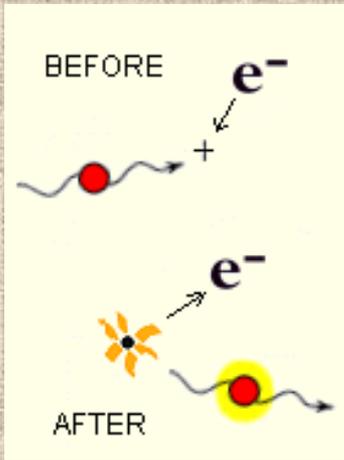
The “NèelCam”, with the **present IRAM optics**, could take **up to 6'** (requires to reduce a bit the IRAM focal length for fitting the 6' on the existing, small M3)

That will require, for Nyquist at 2 mm, a **48×48 (2000 pixels) array**. If the FFTS read-out is working, feasible with a total of 4 coax cables.



BUT ... we'll see ...

S-Z in clusters of galaxies

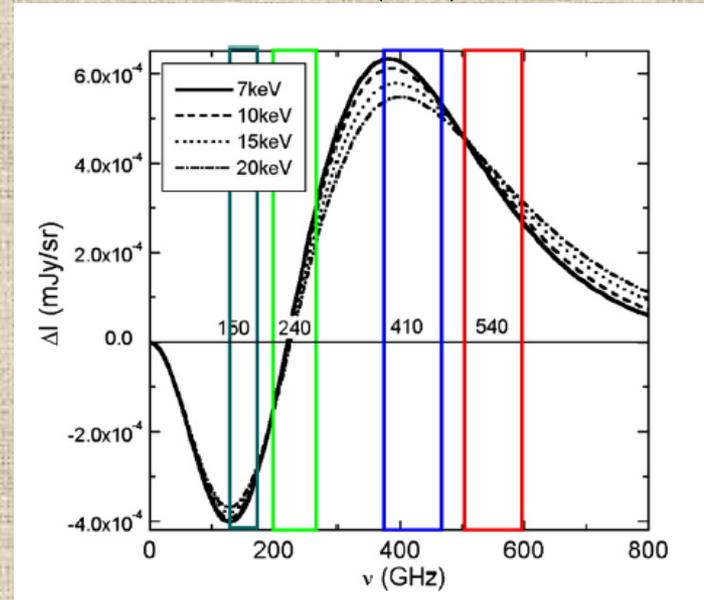
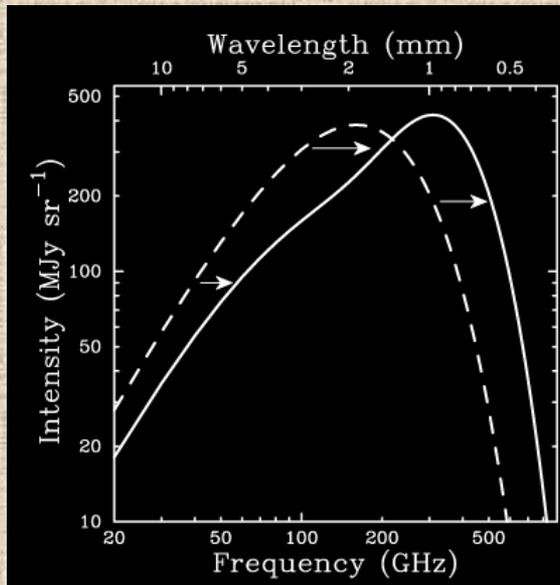


Electrons are really hot:

$$k_B T_e \approx \frac{GMm_p}{2R_{\text{eff}}} \approx 7 (M/3 \times 10^{14} M_\odot) (R_{\text{eff}}/\text{Mpc})^{-1} \text{ keV}$$

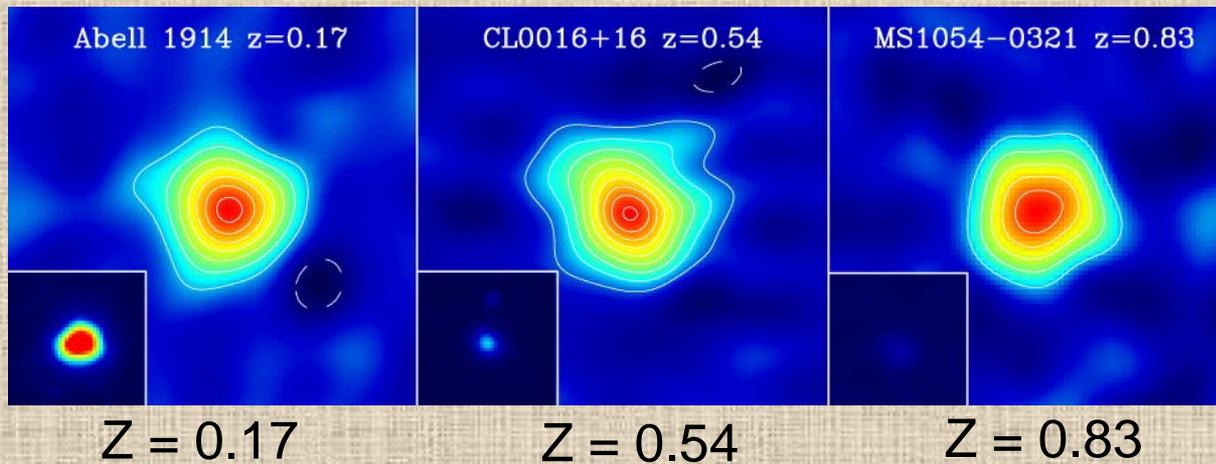
Basic mechanism:

Compton scattering on hot electrons (IC)



S-Z for the deep Universe

SZ contours are $0.75 \mu\text{K}$ and X-ray scales are the same.



SZ particularly important for deep studies ($z > 1$).
Formation and evolution of clusters, Cosmology.

IRAM 30-m



DATA SHEET ON-A-STAMP

- 30 m dish at Pico Veleta (2850m)
- PSF OK down to 0.8 mm
- Heterodyne and Bolometers
- **MAMBO2** -117 pixels. $f = 240\text{GHz}$
Horn-coupled; HPBW = 11",
pixel spacing = 20".

Brand new **European consortium** for a KIDs 2-mm demonstrator at IRAM:

FRANCE: Grenoble (Nèel, LAOG, IRAM). Cryostat, optics, LEKID, digital readout electronics (32 channels), detectors design and fabrication.

UK: AIG Cardiff. Filters, LEKID, digital electronics.

Holland: SRON Utrecht and Groningen: antenna-coupled KIDs design + fabrication

Italy: La Sapienza Roma: LEKID design, software.

Other Applications:

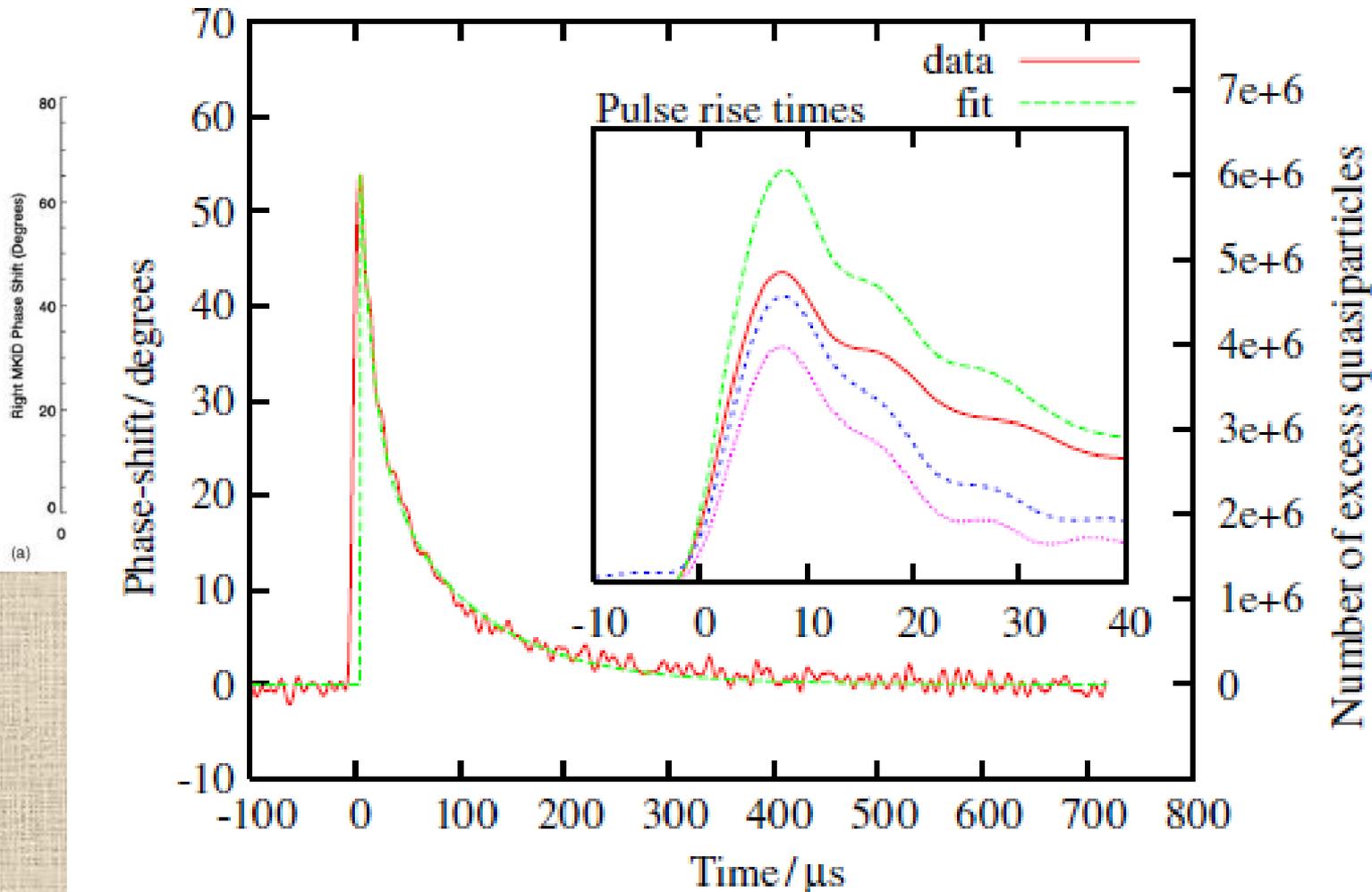
High energy resolution X-rays

Optical/NIR

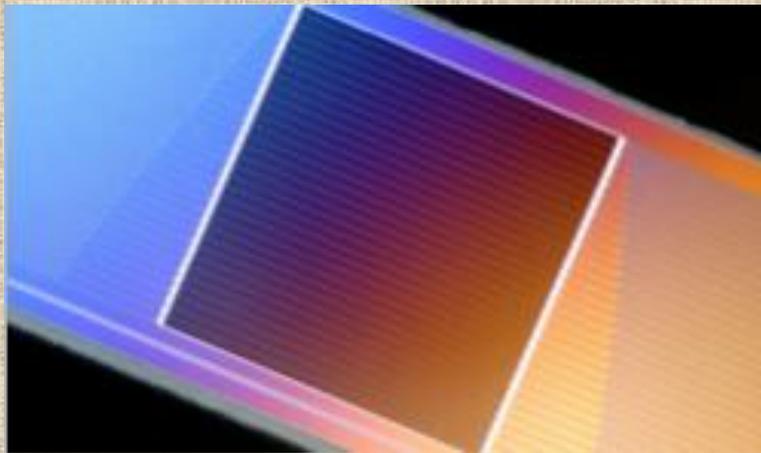
Neutrino mass

.....

X-rays (Caltech, Cambridge)



Optical Camera (UCSB)



From B. Mazin website

An Optical/UV Camera for the Palomar 200" Telescope

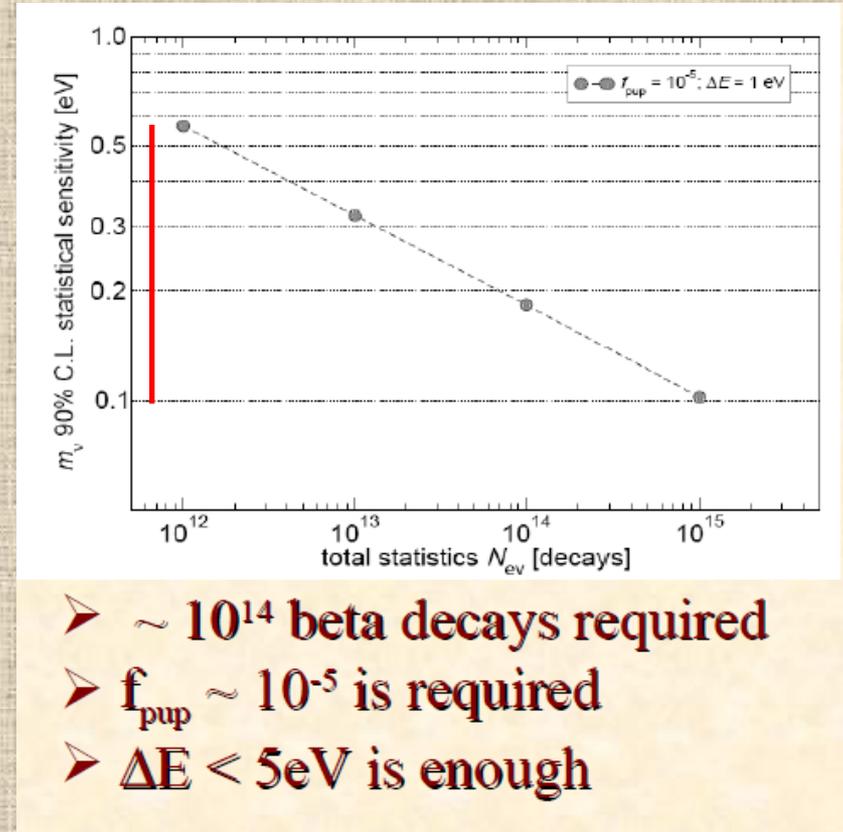
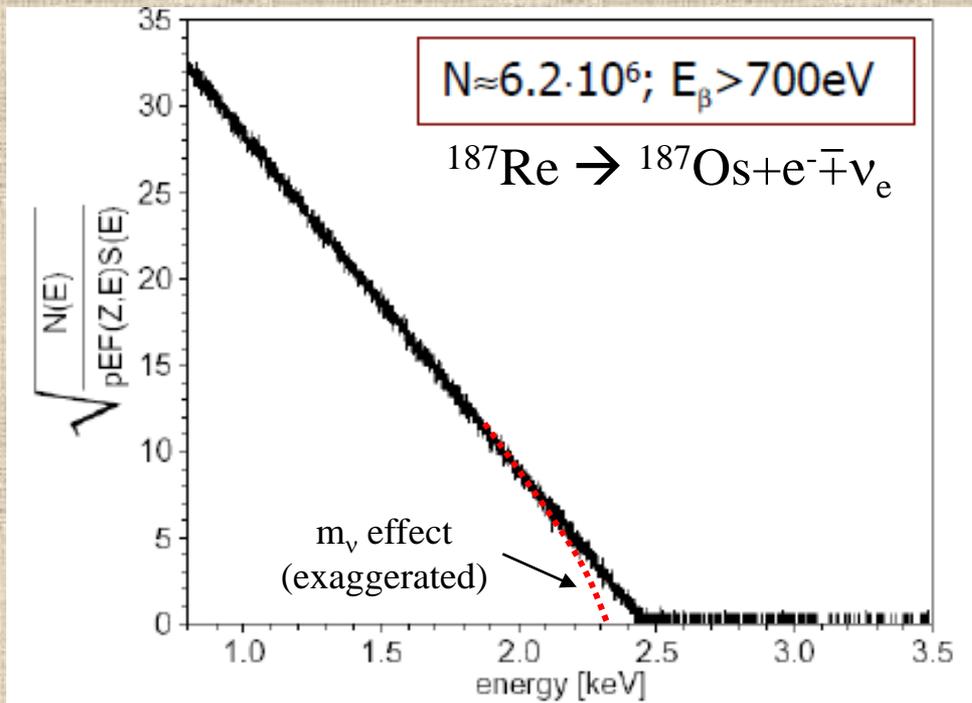
GOAL Build a MKID-based photon counting, imaging spectrophotometer with 64x20 pixels, an energy resolution $R \sim 20$, and 5 microsecond time resolution. This project is a rare example of an extremely exciting instrument that is small enough for a students and post-docs to play an integral part.

STATUS Final array development. Expected first light in December 2009.

Possible science case: GRB fast follow-up. Competing with AMICI prisms + CCD instruments (e.g. REM, ...).

ADVANTAGE: photon-counting

MARE – Microcalorimeters Arrays for a Rhenium Experiment



Growing interest in Grenoble to develop high-quality Re resonators for MARE and q-bits. Epitaxial Re on Shapphire → hopefully very high Q (in progress)

SOME NEW IDEAS FOR MARE (version « resonators » ≠ KIDs)

WE'RE SLOW DUE TO NIKA ... HOPEFULLY BETTER FROM SEPTEMBER 2009 ...

